

TWM/TPQA and Algorithms

Traceability Routes for Electrical Power Quality Measurement

27. – 28. May 2019

ČMI, Brno

Presenter: Stanislav Mašláň, Czech Metrology Institute

Traceability Routes for Electrical Power Quality Measurement

EMPIR Research Potential project

Project code: 15RPT04 TracePQM

Running time: June 2016 to May 2019

Consortium: 13 Partners + 1 RMG

7 already **developed** capability

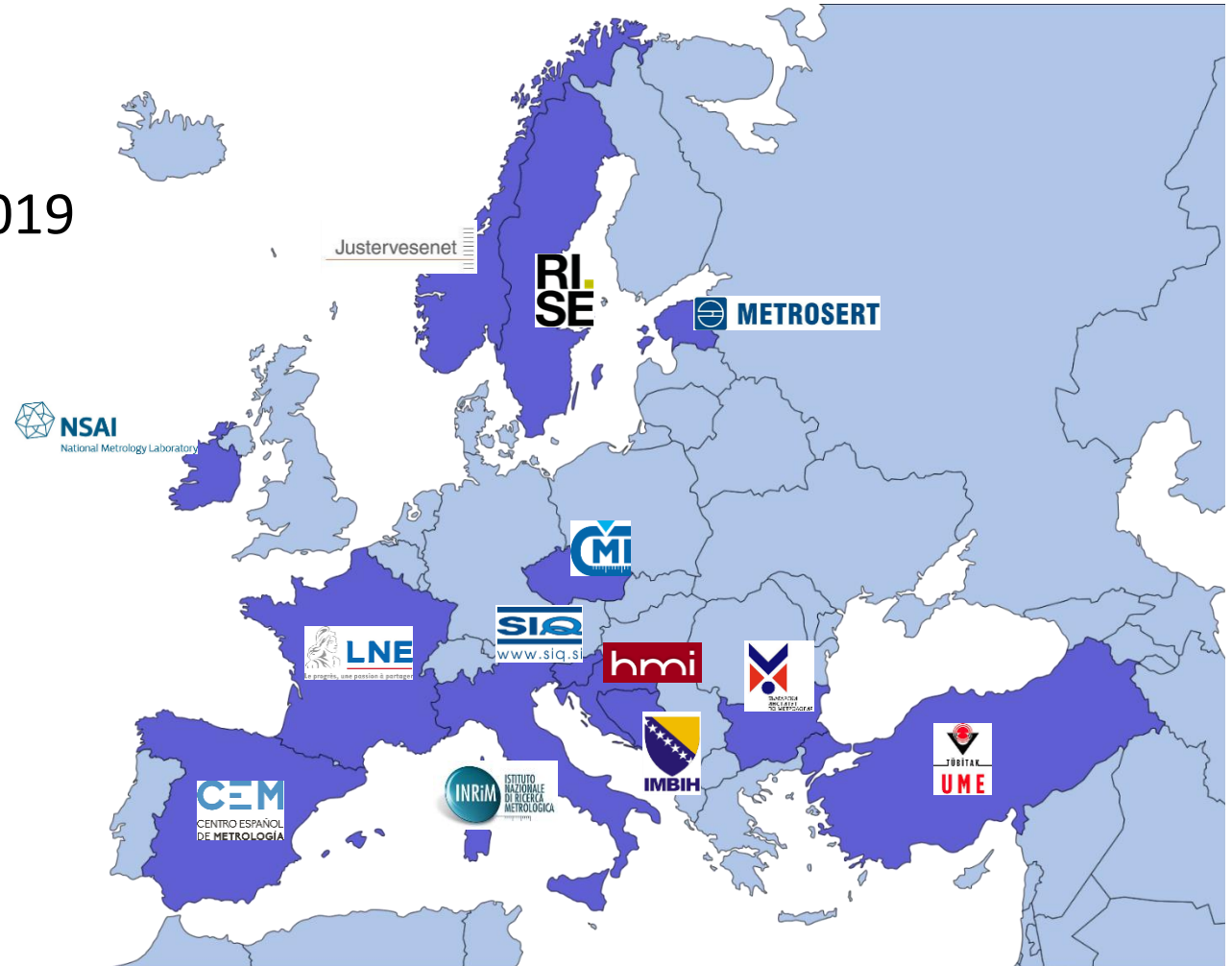
2 in **process** of development

2 intend to **develop**

Potential to **knowledge transfer**

Coordinator:

Věra Nováková Zachovalová



Why Power Quality (PQ)?



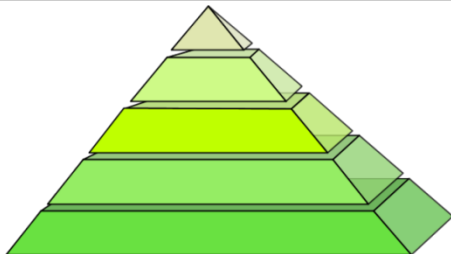
Power Distribution Grid

- Many small “green” sources with **fluctuating power** in the grid
- Non-linear loads (appliances) - **distortion**
- Damage of highly sensitive electronic equipment
- In worst case risk of **blackouts**



Power Quality Monitoring

- Large number of diverse parameters (harmonics, flicker, dips, swells, ...)
- IEC 61000-4-30, IEC 62586-2, IEC 61000-4-15, IEC 61000-4-7
- Wide range of measuring instrumentation



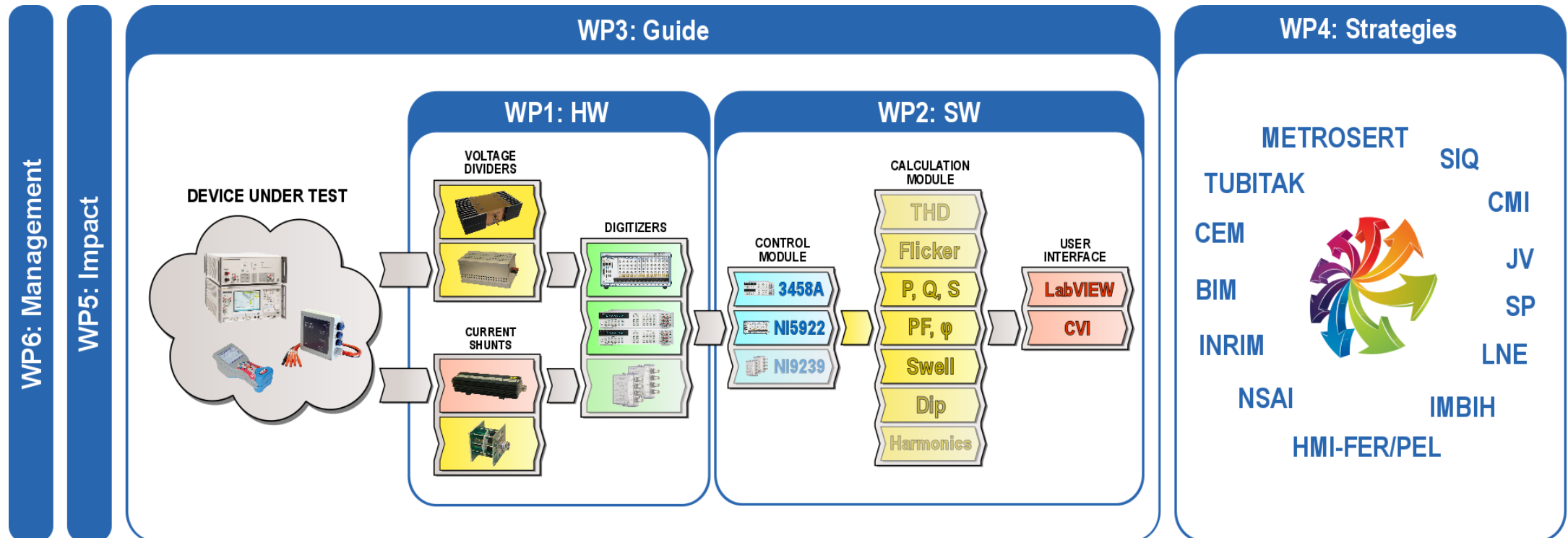
**SI Traceability of PQ measurements is imperative
to ensure comparability of results**

BUT

Difficult to obtain traceability for all PQ parameters

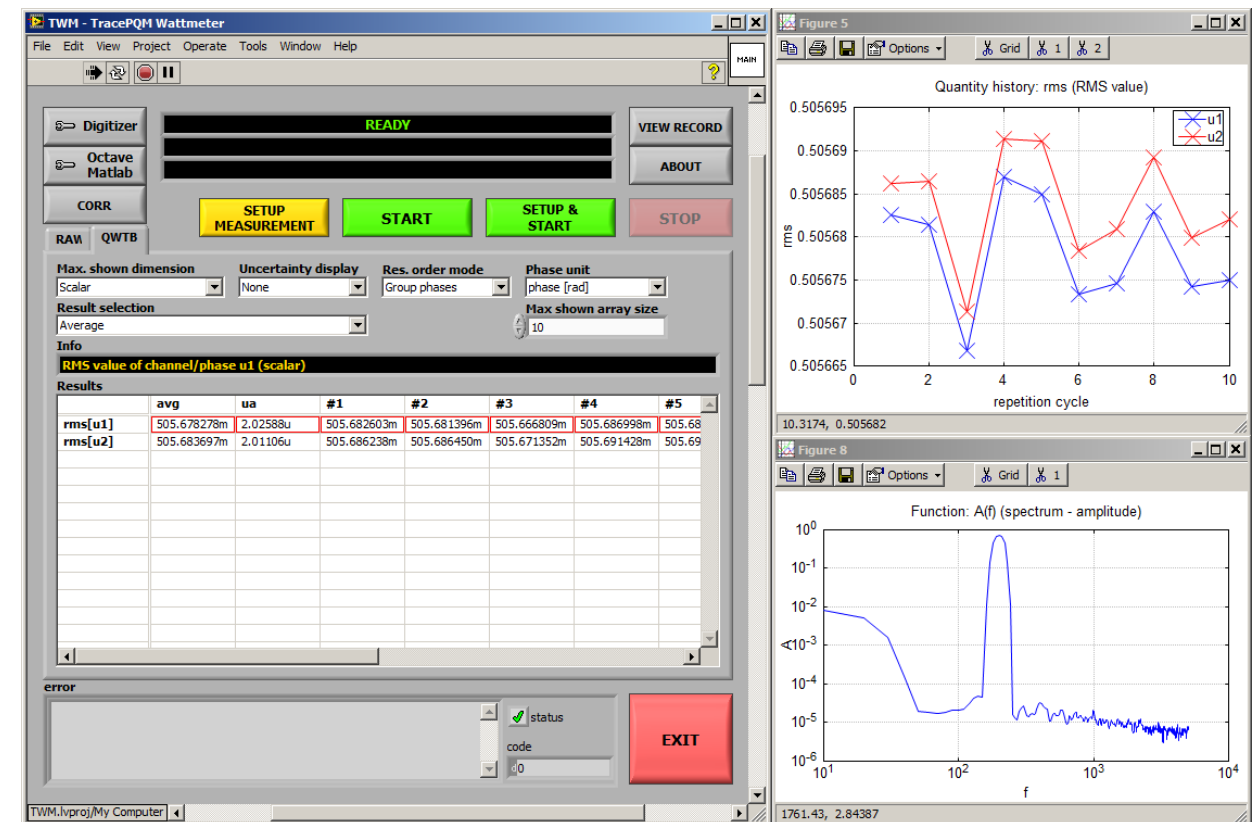
Project structure

Overall goal: To **develop** and **validate** an **open, modular** and well **documented metrology grade system** for sampled power and PQ parameters measurements, which can be easily established at all NMIs and other interested parties.



Open software tools - TWM

- Digitize voltage/current waveforms using:
 - Keysight 3458A DMMs
 - NI PXI 5922 wideband digitizers (niScope compatible)
 - SoundCard (testing purposes)
 - Future: Oscilloscopes, QuADC quantum digitizer, ...
- Any number of digitizer channels
- Save waveform data in unified format
- Load **correction** for each component of setup
- **Execute** selected **algorithm** on waveform data to obtain desired **power** or **PQ parameter(s)**
- Display result as table or graph
- Export results
- **Archive data and results for future processing**



- Written in **LabVIEW** + **Matlab/Octave**
- Full open source: <https://github.com/smaslan/TWM>

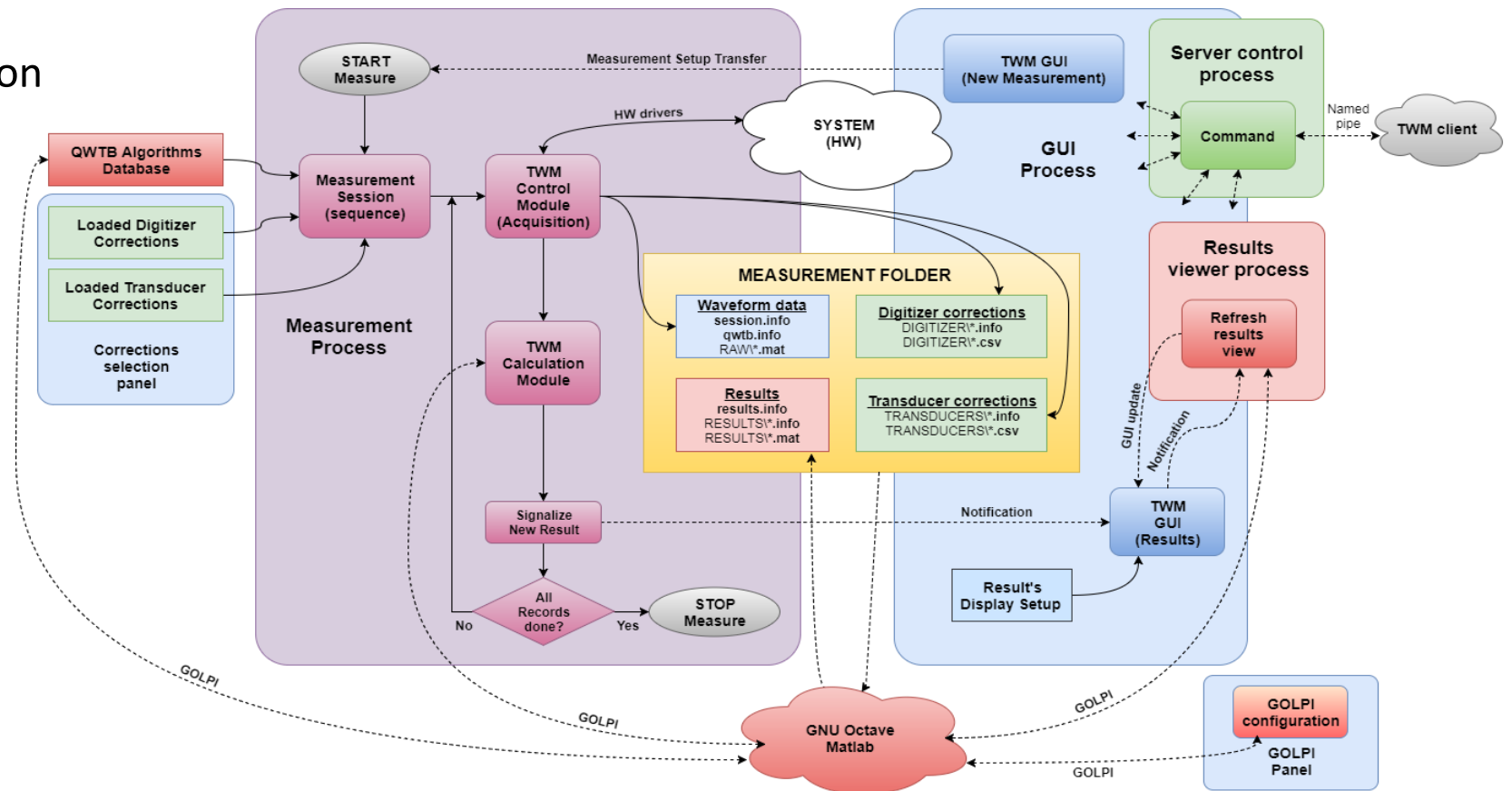
TWM Concept

Two components:

- Control module in LabVIEW
- Processing in Matlab/Octave
- Interlinked to single application using GOLPI

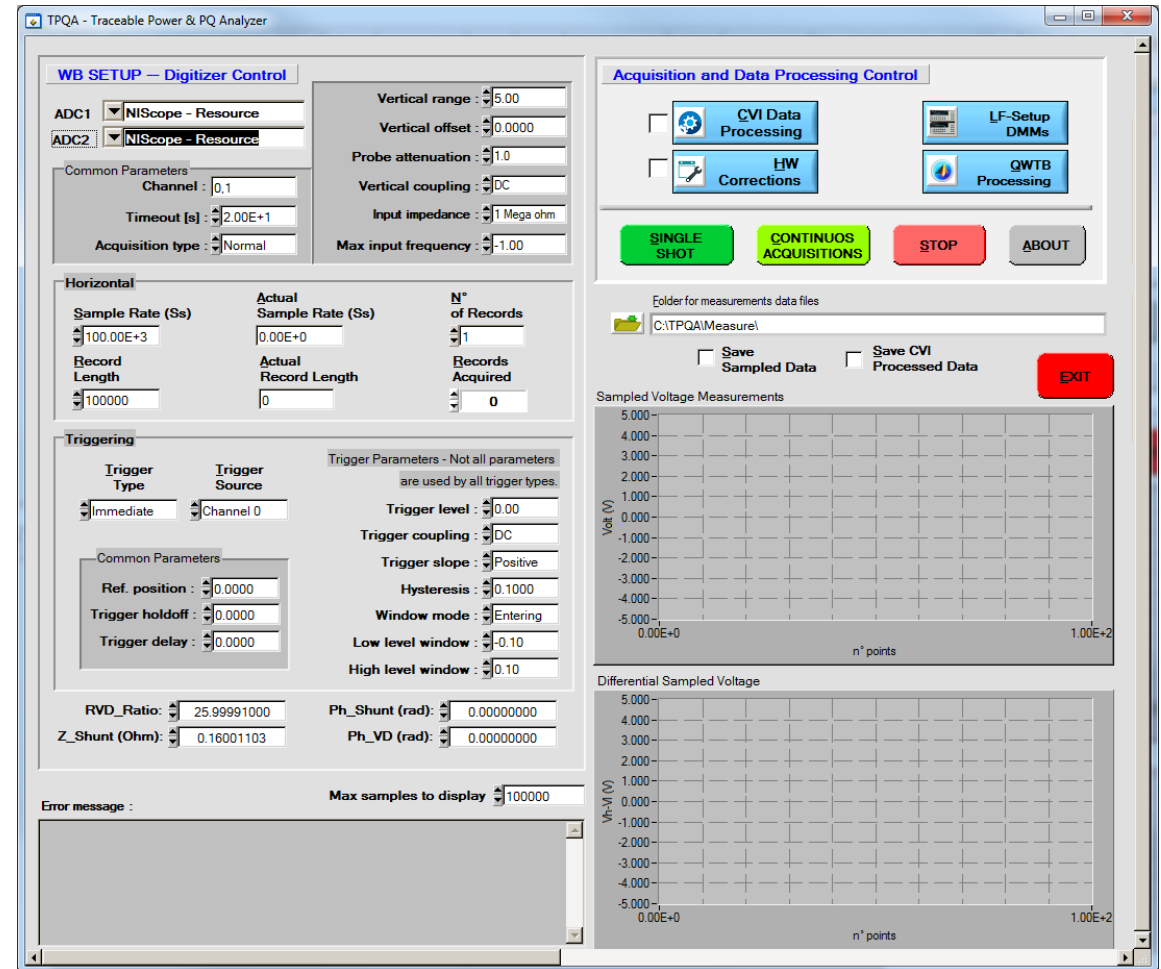
Both can operate standalone:

- Digitize without processing
- Batch processing later
- Reprocessing the raw data for **new parameters**
- Computationally expensive **uncertainty evaluation on supercomputer**
- Validation of algorithms
- New algorithms needs just a few m-files to be added



Open software tools - TPQA

- Functional equivalent of TWM
- Uses either internal processing or TWM algorithms
- Written in LabWindows/CVI (ANSI C)
- Open source
- Full open source:
<https://github.com/btrinchera/TPQA>



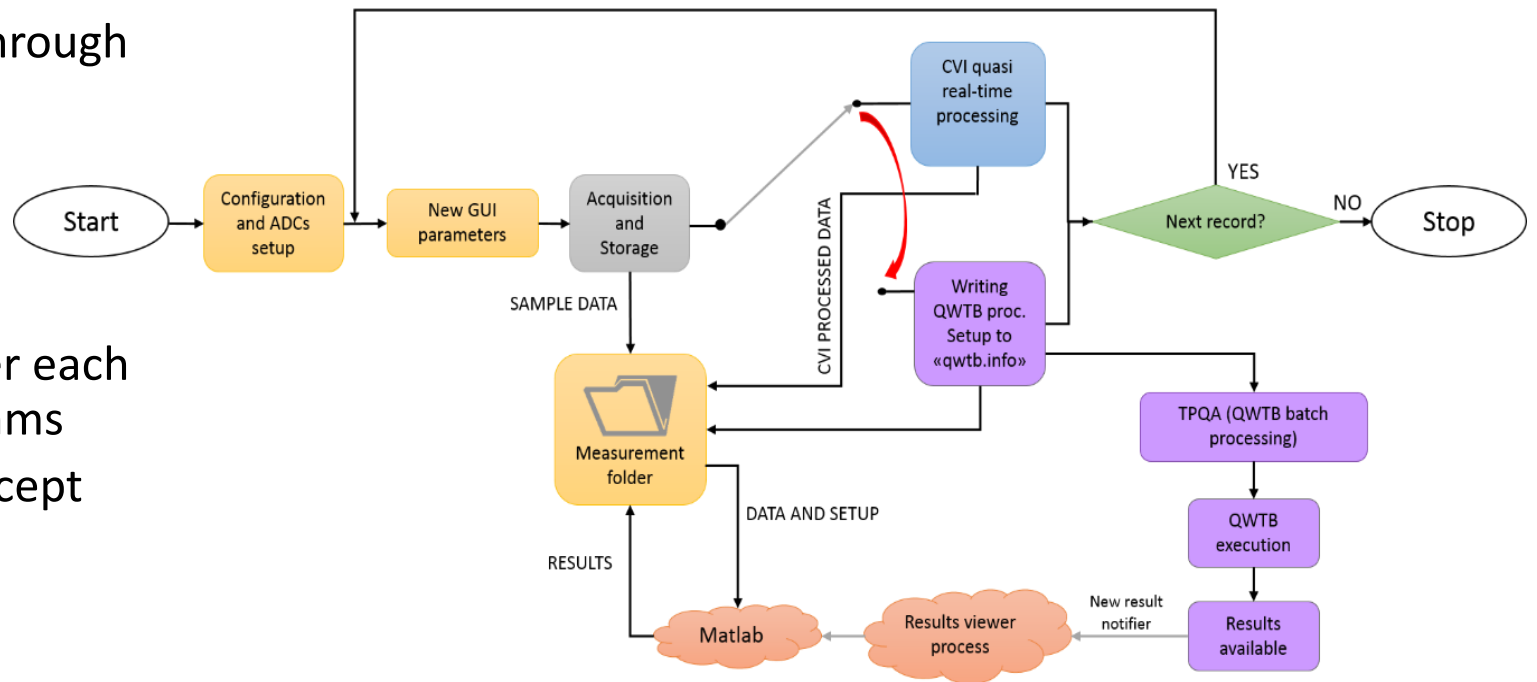
TPQA - Concept

Two components:

- Control module in LabWindows/CVI
- Processing in Matlab and CVI native algorithms
- Matlab linked Labwindows/CVI through matlab runtime library.

Both can operate standalone:

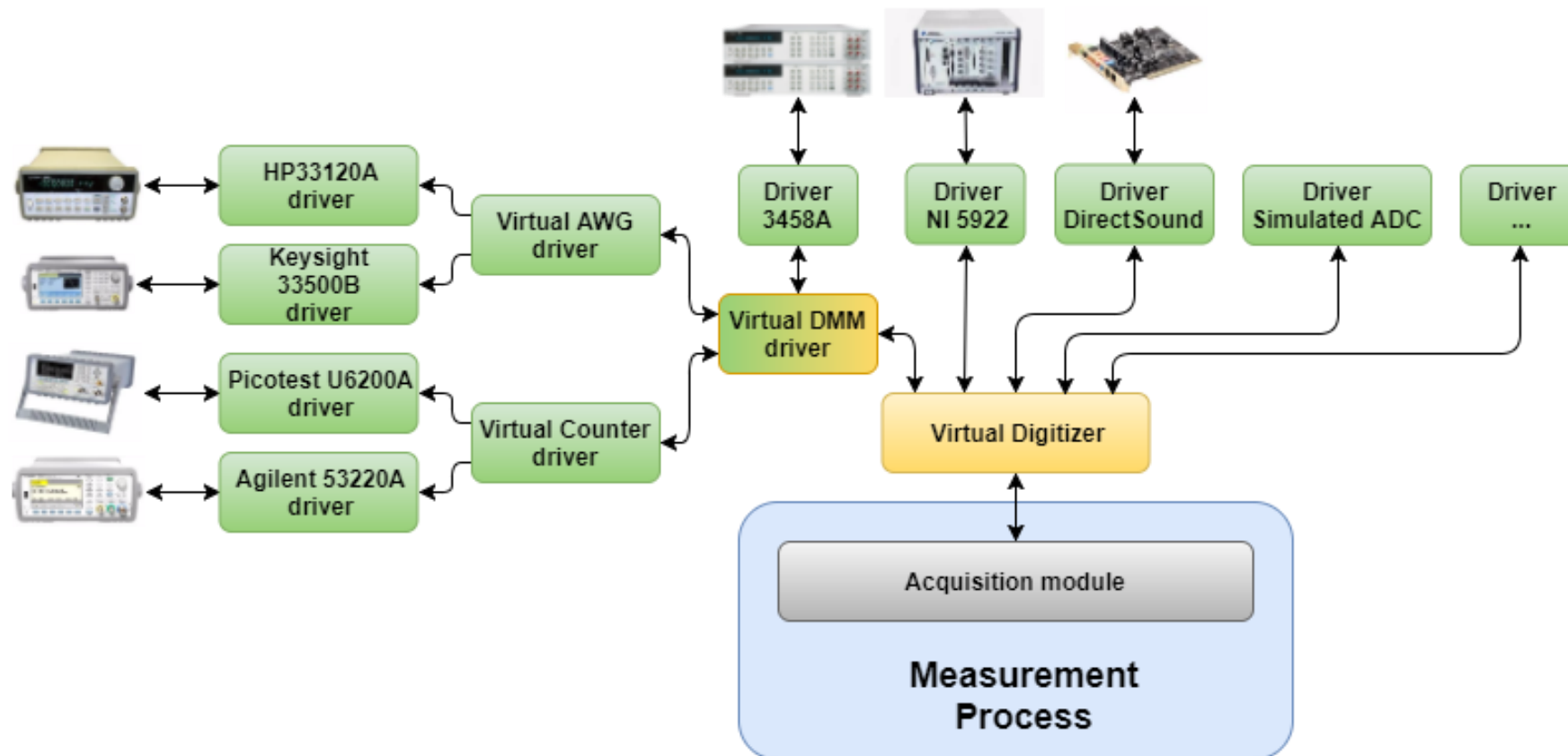
- Digitize without processing
- Data processing immediately after each acquisition using CVI algorithms
- Batch processing using TWM concept
- Reprocessing the raw data with new algorithms
- New algorithms needs just a few m-files to be added



TWM Concept – Digitizer drivers

Generalized digitizer:

- Acquisition module accesses HW drivers via **virtual layer** – same access to any digitizer
- Easy addition of new digitizer – just adding translation from HW drivers to Acquisition module



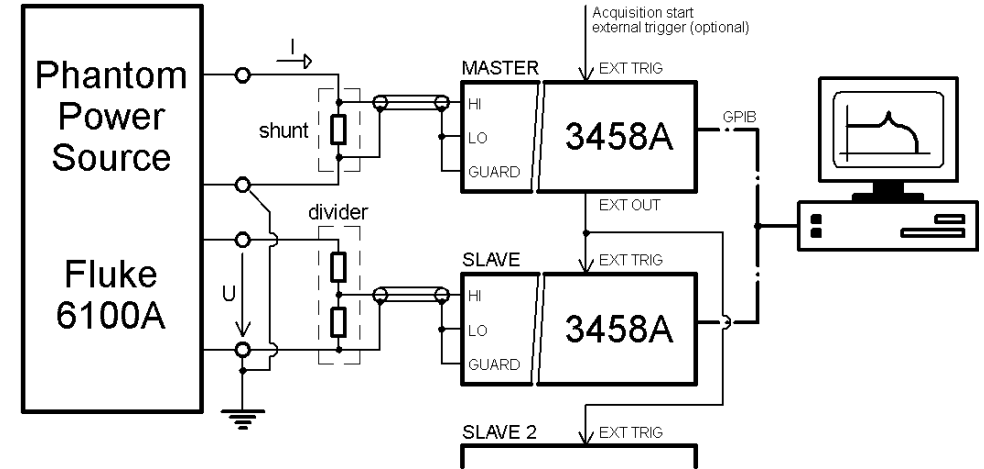
TWM concept – Sampling with Keysight 3458A

Features:

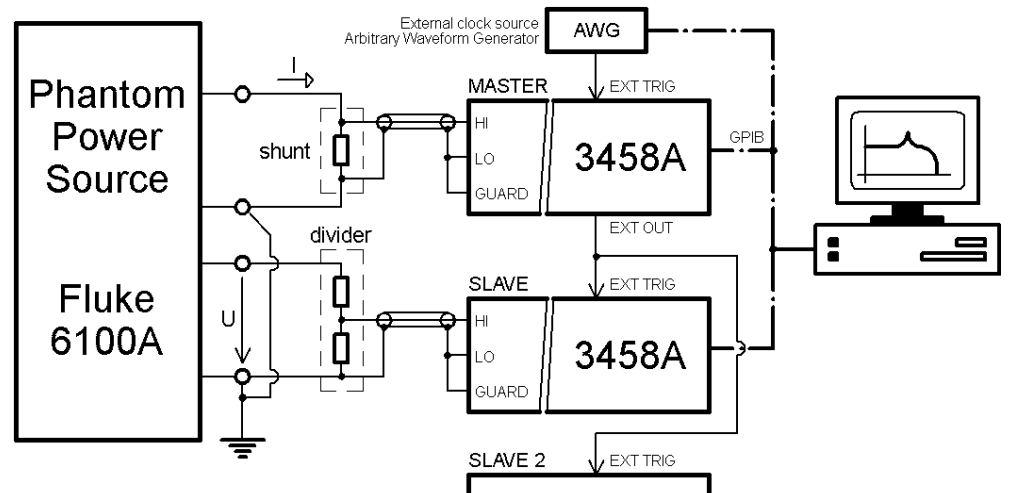
- Simultaneous sampling at **any number of channels**
- Sampling rate **up to 100 kSa/s**
- Memory mode (limited by internal RAM)
- Streaming mode (**up to 16 MSamples** per channel)
- Modes: **DCV** or DSDC, DSAC
- Sample clock modes:
 - MASTER-SLAVE, internal clocking (**non-coherent sampling**)
 - MASTER-SLAVE, external clock source
 - Modified HW, locked to 10 MHz (**experimental**)
- Trigger:
 - Immediate
 - Level
 - External



Internal clocking by MASTER:



External clocking from AWG or another pulse clock:

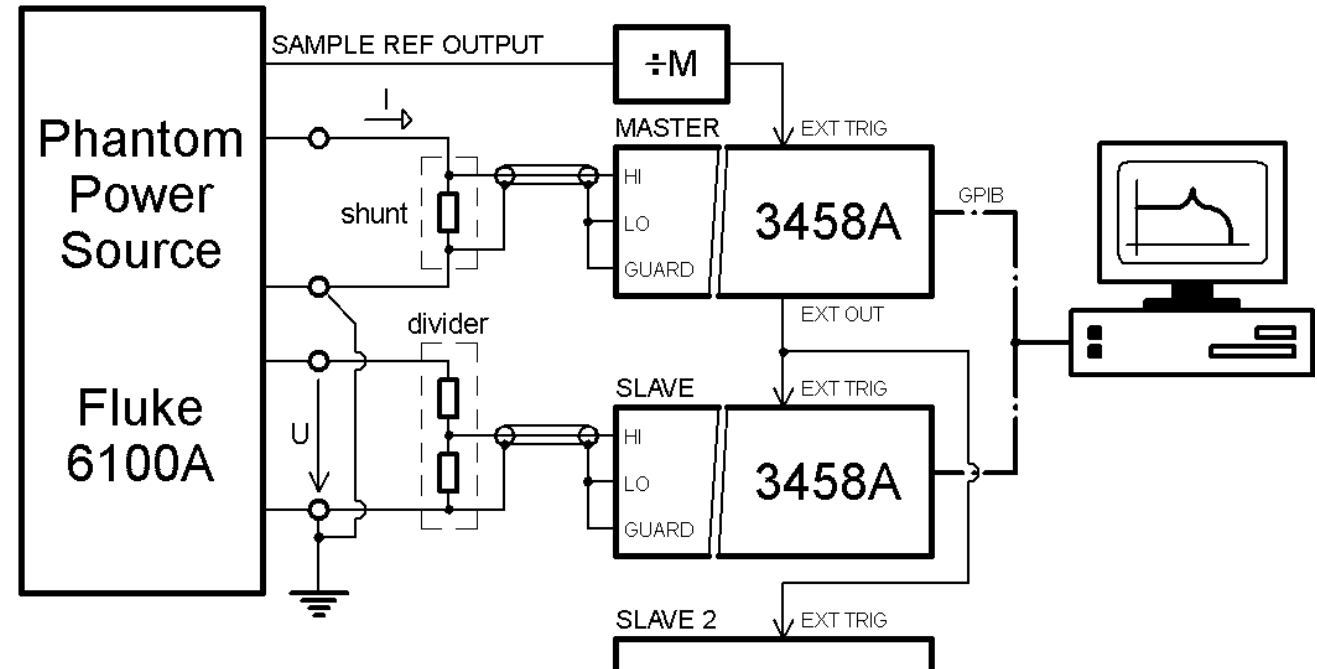


TWM concept – Sampling with Keysight 3458A

Special case:

- Sample clock taken from source
- Fluke 6100A outputs:
 - sample clock out = $f_0 * N$
 - Must be enabled by command:
OUTP:ROSC ON
- Useful for coherent sampling:
 - Integer number of signal periods per record
- Problem:
 - 6100A cannot control the **N** factor.
It is set so sampling is always **> 50 kHz =>**
aperture of 3458A must be short =>
high noise and gain errors!

External sample clock from source:



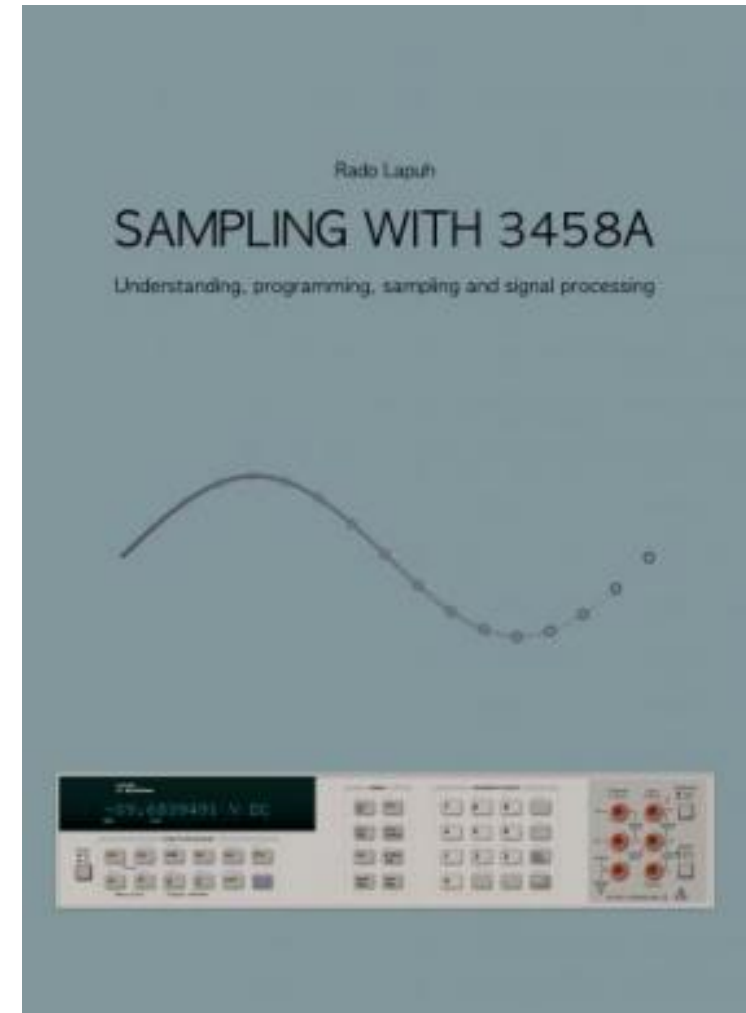
TWM concept – Sampling with Keysight 3458A

Highly recommended literature:

Excellent book from Rado Lapuh on 3458A:

- Sampling,
- Processing data,
- Connections,
- Special features
- Characterisation

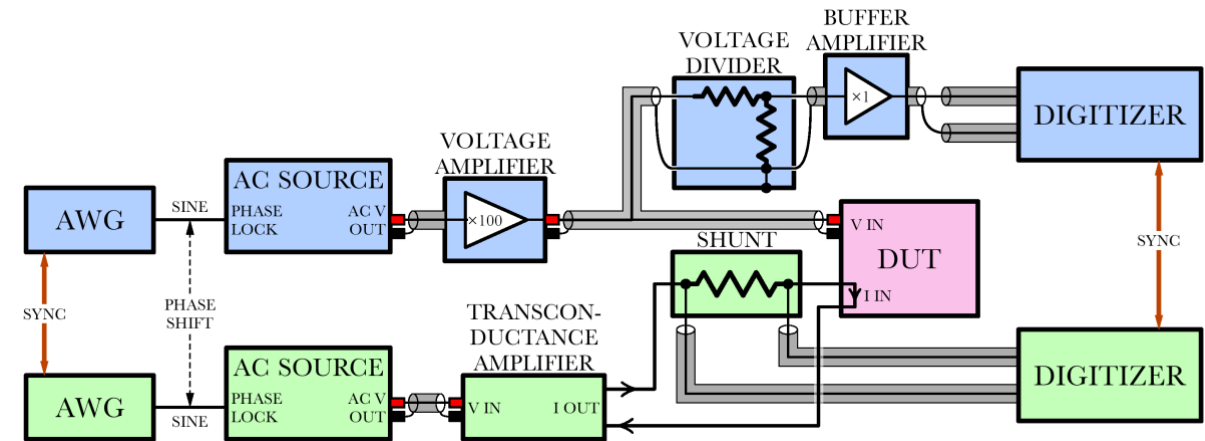
Source: **Rado Lapuh, “Sampling with 3458A”,
ISBN 978-961-94476-0-4**



TWM concept – Sampling NI 5922 digitizers

Features :

- Any number of channels
- Coupling: DC, AC
- Input impedance: 1 M Ω or 50 Ω
- Connection mode: single-ended, differential
- Memory mode (up to board limit)
- Streaming mode (unlimited sample count)
- Multi-board synchronization (niTClk)
- Optional lock to external 10 MHz
- Trigger mode:
 - Immediate
 - Level
 - External



Wideband setup, with 5922 (RISE)

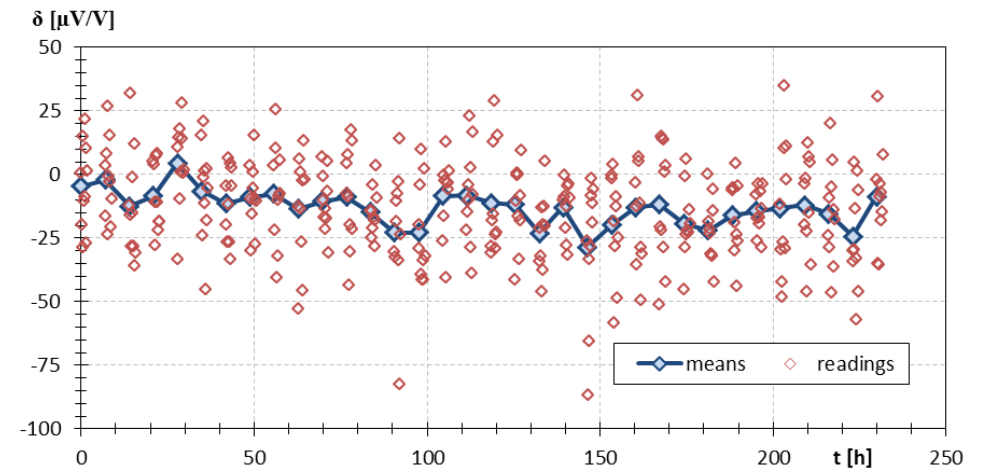


TWM concept – Sampling with NI 5922 digitizers

Features :

- Sampling rate 50 kSa/s to 15 MSa/s
- Unlike 3458 no aperture setting
- Resolution varies from 24 bit down to 16 bit depending on the sampling frequency
- Limited gain stability
- Temperature dependence: up to some 20 ppm/°C (internal temperature)
- Limited repeatability of self-calibration: up to ± 90 ppm!

Repeatability of gain for repeated self-cal:



TWM concept – Sampling with SoundCard

Features :

- Experimental only!
- **One** or **two** channels
- Sampling rate limited by HW to some 192 kSa/s
- Typically high crosstalk
- Usable for e.g.:
 - Phase shift
 - Amplitude ratios
 - THD



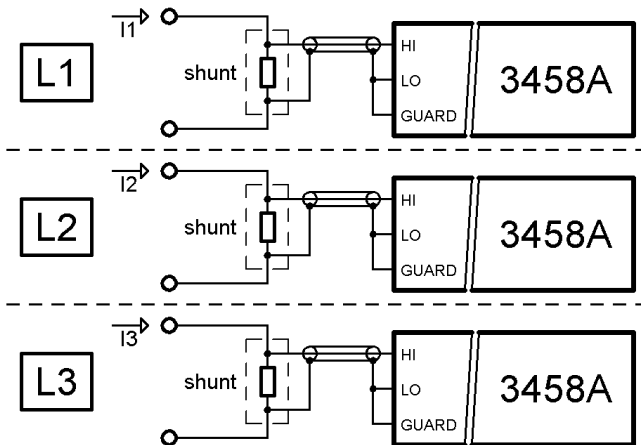
Source: [Creative Technology Ltd.](#)

TWM concept – versatility of transducer configuration

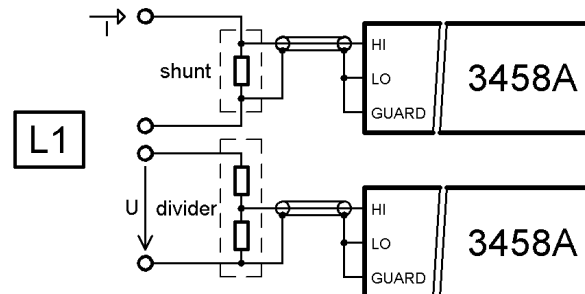
Designed to be able to operate:

- Single channel (single voltage or current input)
- Single phase operation (one voltage and one current input)
- Multiphase operation (multiple voltage and current inputs)

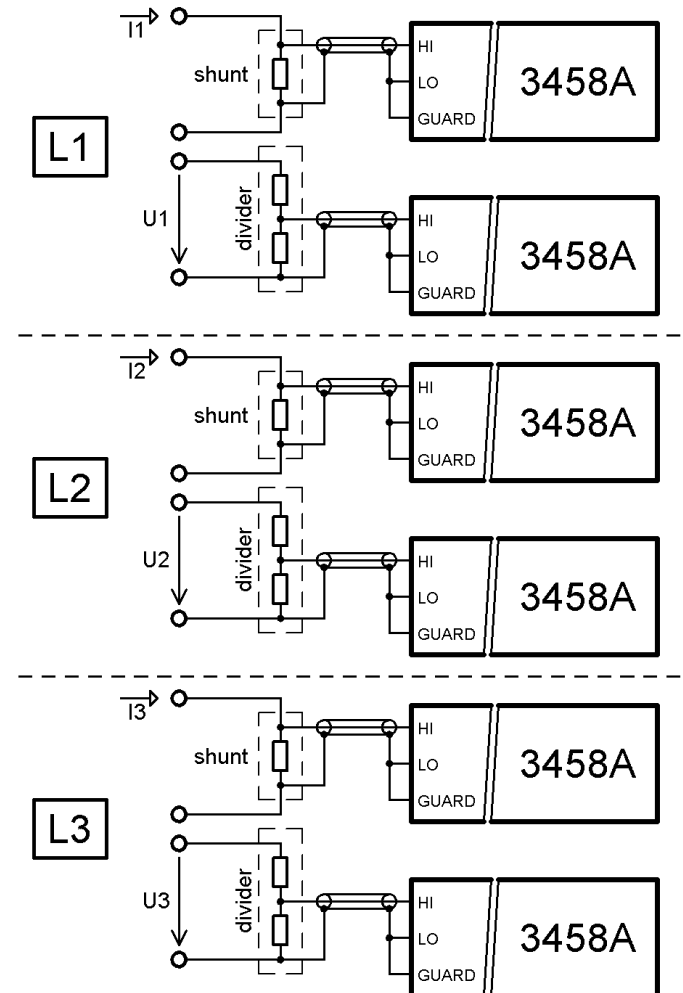
Single channel, multiphase operation:



Single phase operation (power):



Multiphase operation (power):

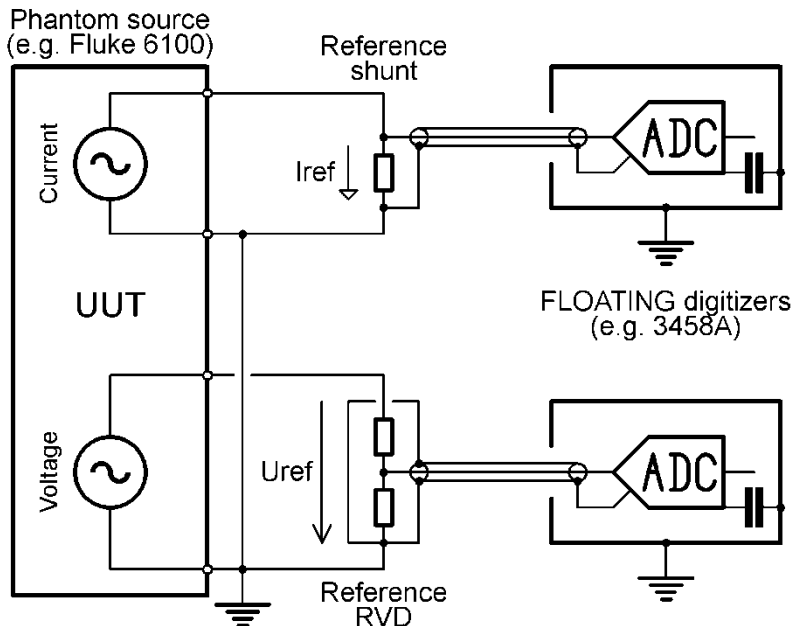


TWM concept – versatility of transducer configuration

Transducer configuration:

- **Single-ended connection** (low frequency)
- Differential connections (wideband operation)
- Mixed single-ended and differential

Single ended transducer connection (LF/HF):

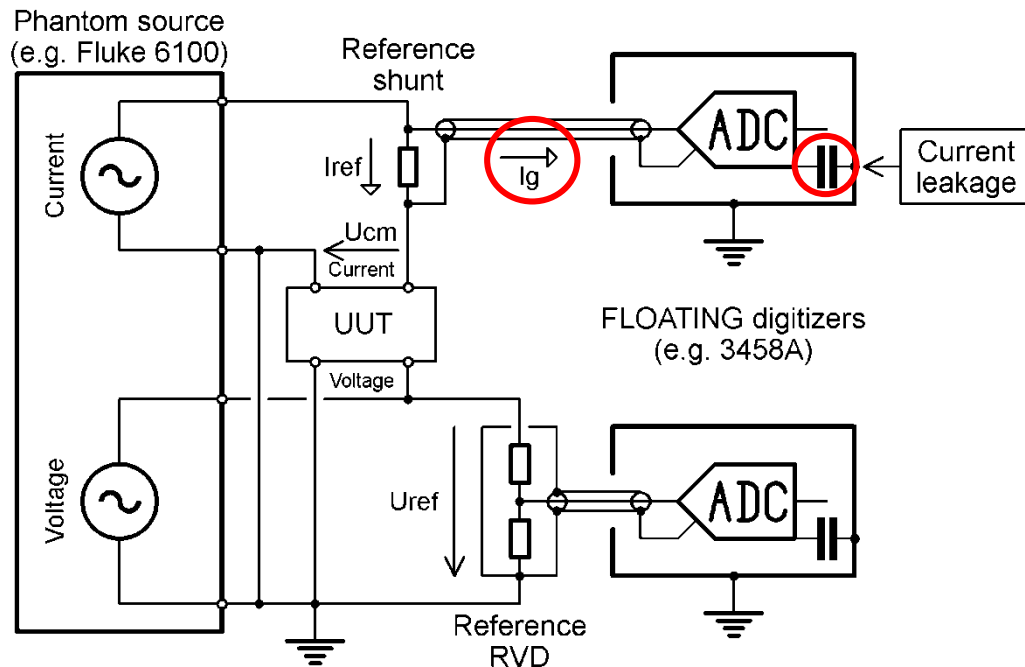


TWM concept – versatility of transducer configuration

Transducer configuration:

- Single-ended connection (low frequency)
- Differential connections (wideband operation)
- Mixed single-ended and differential

Single ended transducer connection (usually LF):

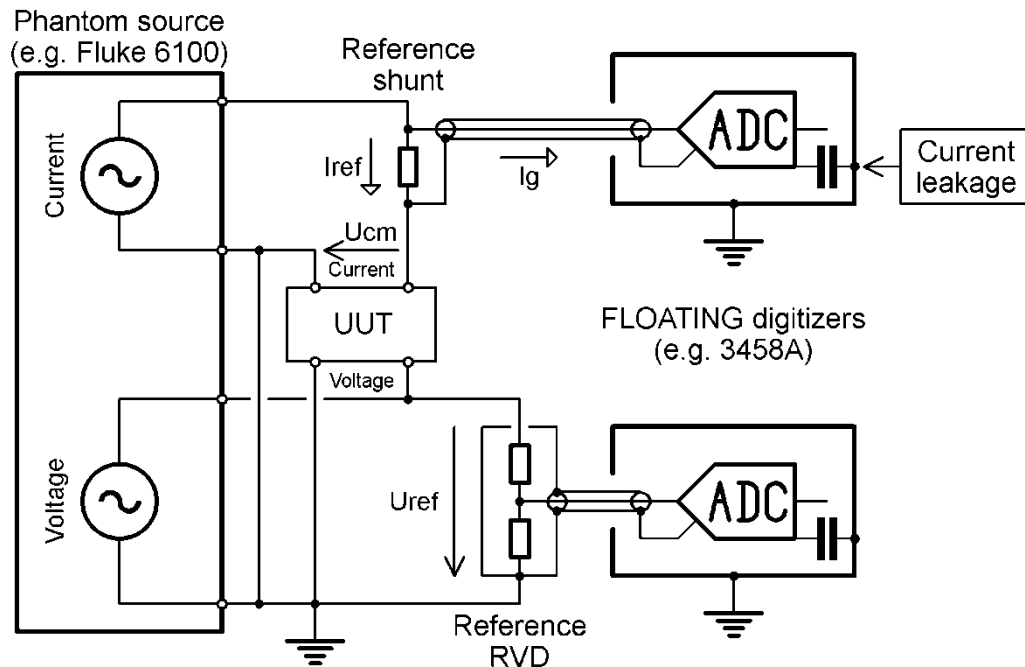


TWM concept – versatility of transducer configuration

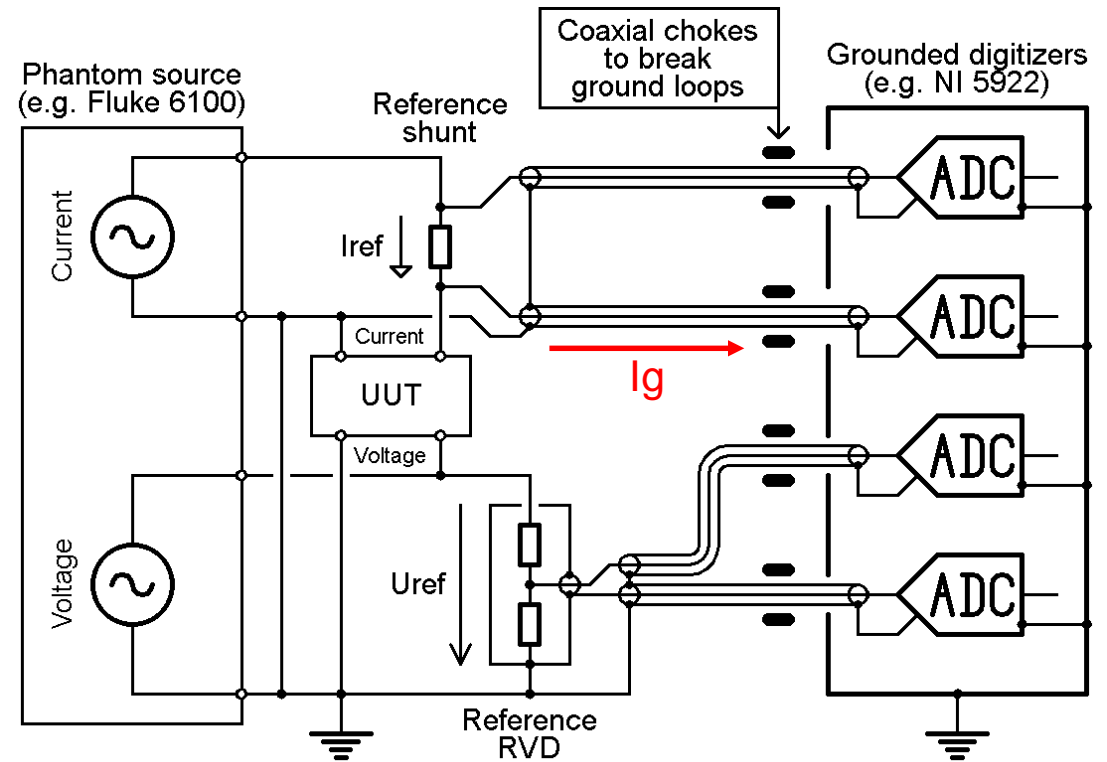
Transducer configuration:

- Single-ended connection (low frequency)
- **Differential connections (wideband operation)**
- Mixed single-ended and differential

Single ended transducer connection (usually LF):



Differential transducer connection (wideband):

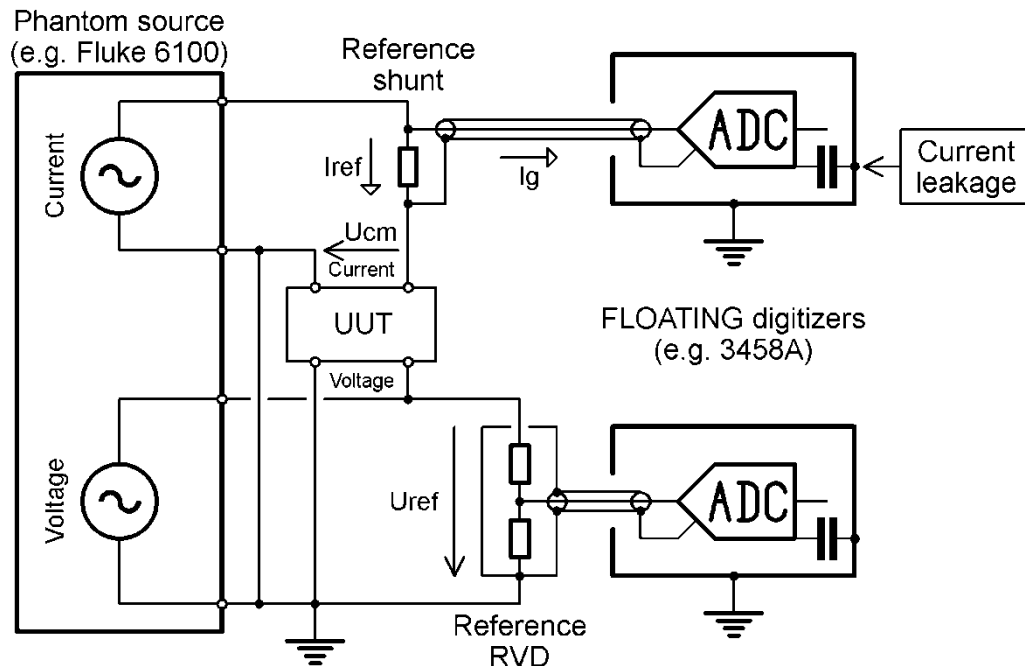


TWM concept – versatility of transducer configuration

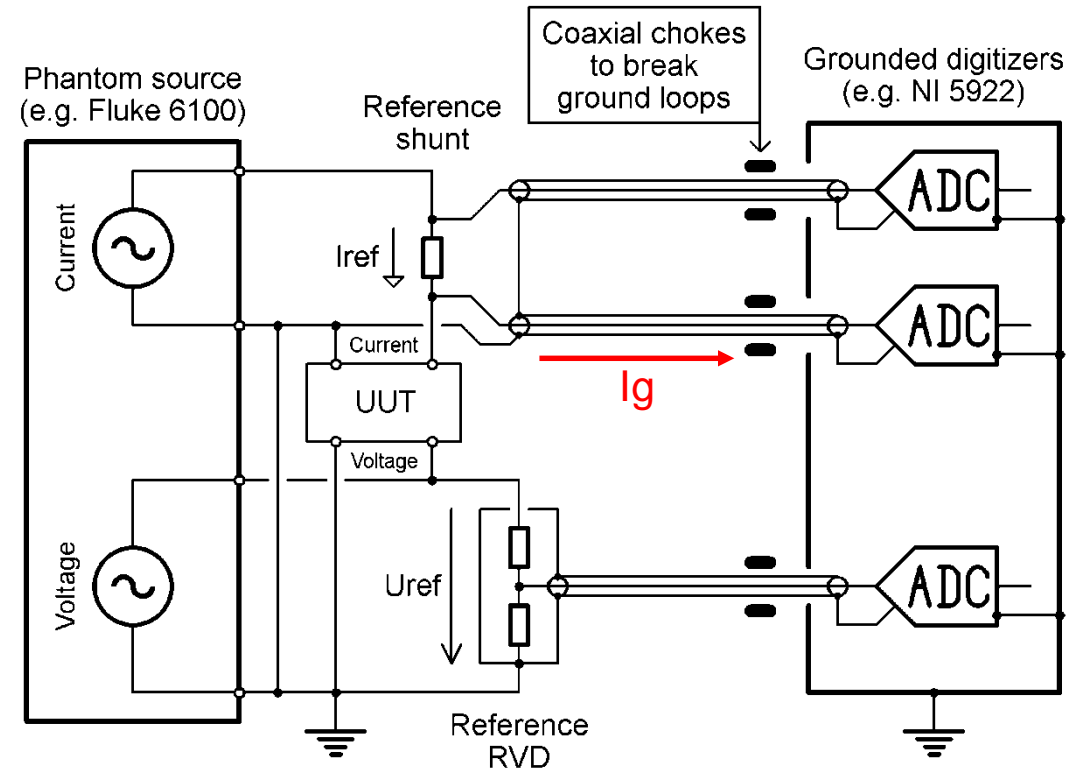
Transducer configuration:

- Single-ended connection (low frequency)
- Differential connections (wideband operation)
- **Mixed single-ended and differential**

Single ended transducer connection (usually LF):



Mixed mode transducer connection (wideband):



TWM concept – Corrections

Features:

- Every component of setup corrected
- All corrections are optional – no need to load any
- All corrections have assigned uncertainty
- Versatile human readable format
- Corrections may be dependent on digitizer attribute
 - Auto selection of various correction files e.g. by temperature or range selected

Types of corrections:

- Digitizer correction
- Transducer corrections
- Cable corrections (part of transducer)

TWM concept – Digitizer corrections

Gain:

- Corrects gain error of digitizer
- Composed of *Nominal gain* and *relative ac-dc* transfer (frequency-amplitude dependent)

Phase:

- Corrects phase shift of channel (frequency-amplitude dependent)
- Can be used to correct inter-channel phase error

Offset:

- Corrects DC offset of digitizer channel

Time-shift:

- Defines time-shift of each digitizer channel relative to first channel

Digitizer aperture:

- Corrects gain and phase error of channel for integration ADC (3458A):

$$gain = \frac{aperture \cdot \pi \cdot f}{\sin(aperture \cdot \pi \cdot f)} [V/V]; phase = aperture \cdot \pi \cdot f [rad]$$

- Often sufficient to replace gain ac-dc and phase for 3458A

TWM concept – Digitizer corrections

Time-base correction:

- Corrects error of digitizer time-base (results in frequency estimation error)
- **Used only for internally generated sample clock, e.g. for 3458A**

SFDR – Spurious Free Dynamic Range:

- Defines spur levels of digitizer channel (frequency-amplitude dependent)
- **It's no correction** – it's used to estimate uncertainty e.g. for THD evaluation

Sampling time jitter:

- **It's no correction** – it's used to estimate uncertainty (increases standard deviation)

Input impedance:

- Defines input impedance of digitizer channel
- It is used as part of loading corrections
- Some digitizers have high input capacitance and low resistance, so this may be essential especially for phase angle

TWM concept – Transducer corrections

Gain:

- Corrects gain error of transducer
- Composed of *Nominal gain* and *relative ac-dc* transfer (frequency-rms dependent)
- Nominal gain is *resistance for shunt*, or *input-to-output ratio for divider*

Phase:

- Corrects phase error of transducer (frequency-rms dependent)

SFDR – Spurious Free Dynamic Range:

- Defines SFDR of transducer (frequency-rms dependent)
- *It's no correction* – it's used to estimate uncertainty e.g. for THD evaluation

RVD low-side impedance:

- Approximate impedance of low-side resistor of resistive voltage divider (part of loading corrections)

Loading corrections:

- Impedances of particular components of connection impedance model

TWM concept – Transducer corrections

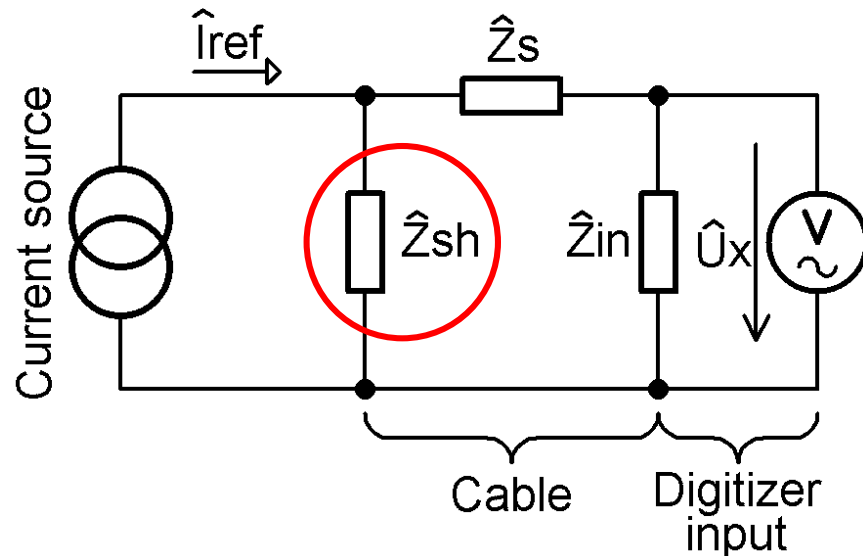
Loading corrections example:

- Shunt impedance Z_{SH} is reduced by parallel impedances ($Z_S + Z_{IN}$) to effective Z_{SH}' :

$$\hat{Z}_{SH}' = \frac{(\hat{Z}_S + \hat{Z}_{IN})\hat{Z}_{SH}}{\hat{Z}_{SH} + \hat{Z}_S + \hat{Z}_{IN}}$$

- Actual measured voltage U_X is connected to the shunt via divider $Z_S:Z_{IN}$ with ratio:

$$\hat{k}_{LOAD} = \frac{\hat{Z}_{IN}}{\hat{Z}_S + \hat{Z}_{IN}}$$



Loading corrected current evaluation:

$$\hat{I}_X = \hat{U}_X \frac{\hat{Z}_{SH} + \hat{Z}_S + \hat{Z}_{IN}}{\hat{Z}_{SH} \cdot \hat{Z}_{IN}}$$

Example:

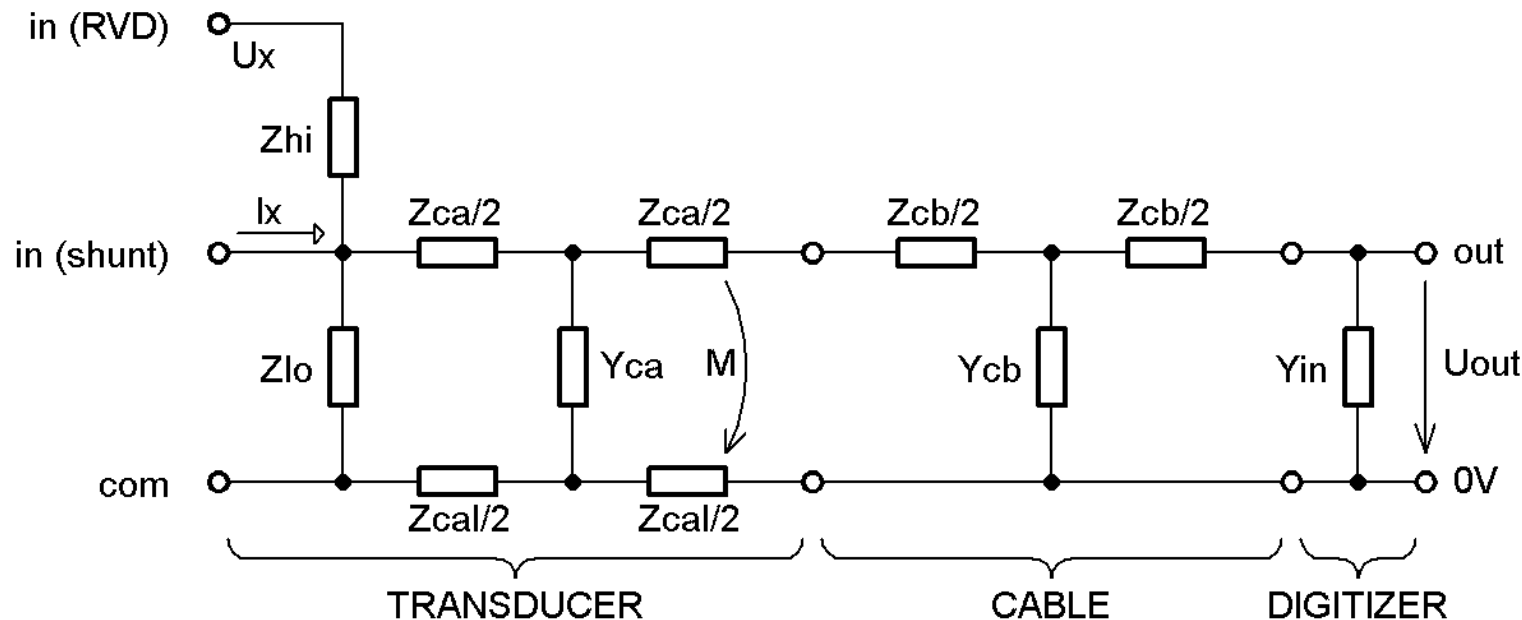
$Z_{SH} = 6 \Omega$, $Z_S = 50 \text{ m}\Omega$, $Z_{IN} = 1 \text{ M}\Omega \parallel 300 \text{ pF}$, $f = 400 \text{ Hz}$
 $\delta I_X = +6.05 \mu\text{A/A}$, $\Delta\varphi = +4.5 \mu\text{rad}$

TWM concept – Transducer corrections

Loading corrections:

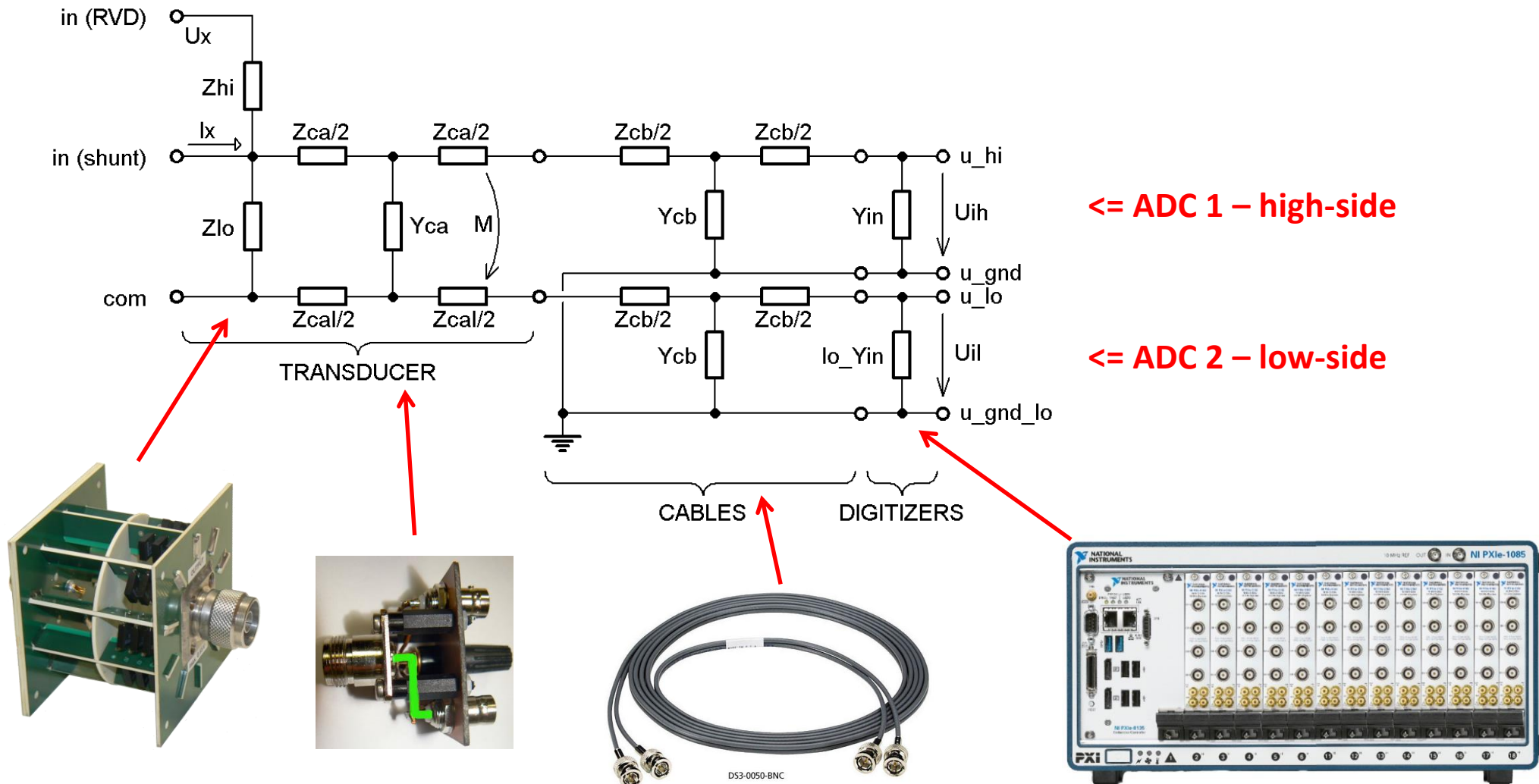
- Corrects effect of cable and digitizer input impedance to transducer transfer
- Requires defined impedance model of interconnections
- Applies mainly at higher frequencies

Single ended transducer loading correction model:



TWM concept – Transducer corrections

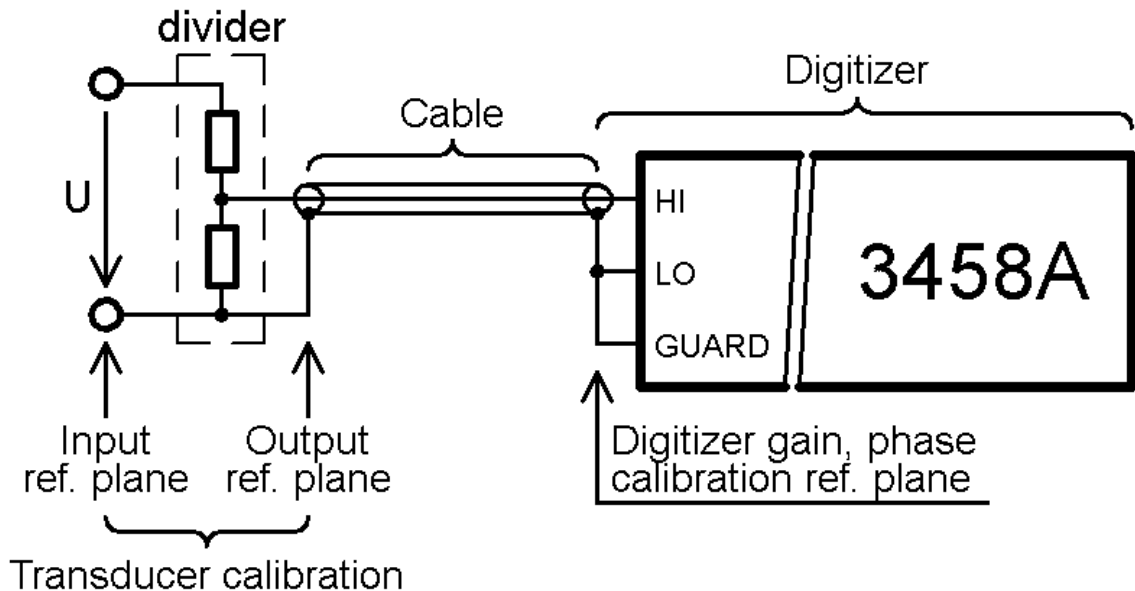
Differential transducer loading correction model:



TWM concept – Calibration concept

Separate **transducer**, **cable** and **digitizer** correction:

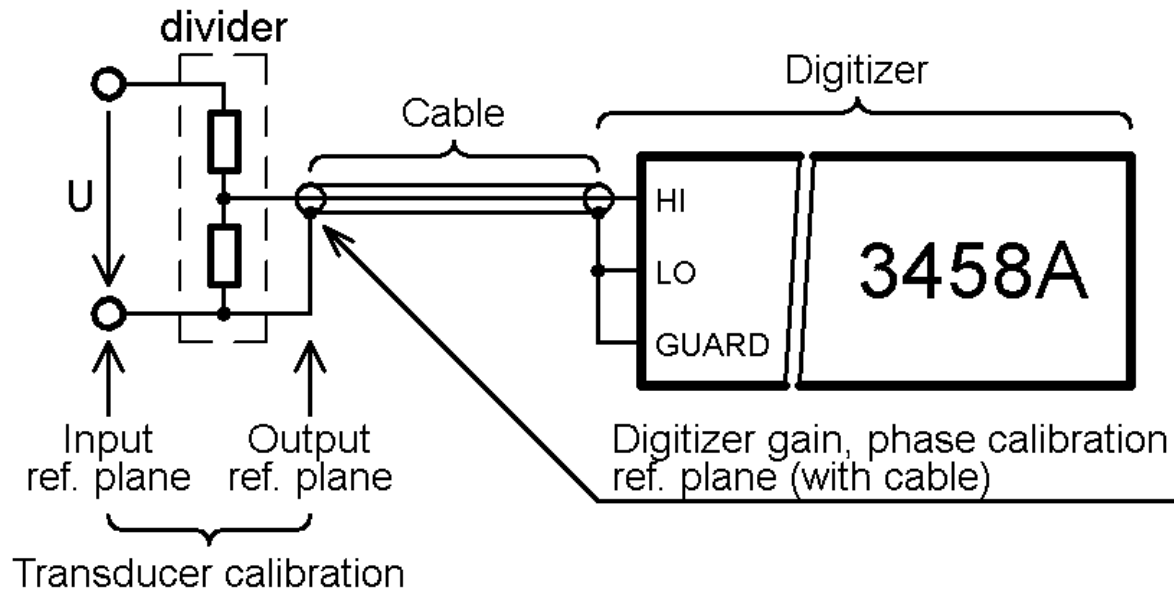
- Allows to use different cables and transducers while digitizer calibration stays unchanged
- **Transducer** calibrated for **no loading** condition (*gain, phase, SFDR*)
- **Digitizer** calibrated **at its terminals** (*gain, phase, offset, SFDR*)
- Cable parameters must be defined (part of loading correction)



TWM concept – Calibration concept

No cable calibration:

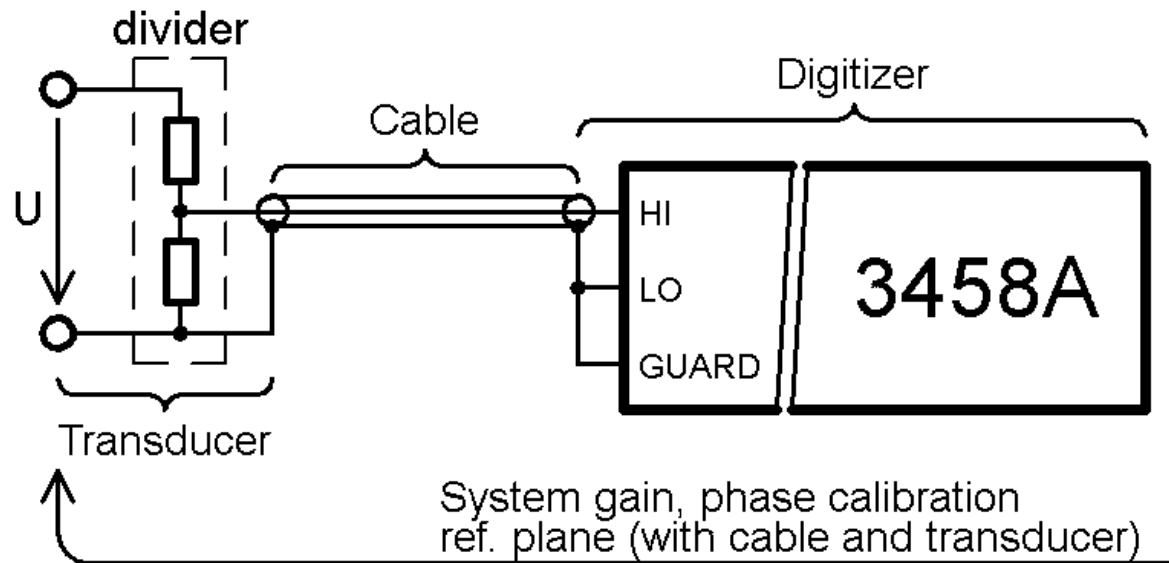
- Digitizer calibrated **together with input cable**
- The **same cable** must be used for **each transducer**
- **Digitizer+cable** gain, phase, offset, SFDR are stored to **digitizer correction**
- **Transducer** gain, phase, SFDR are stored to **transducer correction**
- **Cable** corrections are left **undefined** (no correction)



TWM concept – Calibration concept

System calibration:

- Simple calibration – transducer, cable and digitizer **calibrated together as single device**
- **Total channel** *gain, phase, offset, SFDR* stored to **digitizer correction**
- **Transducer** and **cable** corrections are left **undefined** (no correction)
- Requires **known source** of input voltage/current (calibrator or precision voltmeter, e.g. Fluke 5790A)



TWM Algorithms

Coffee time!

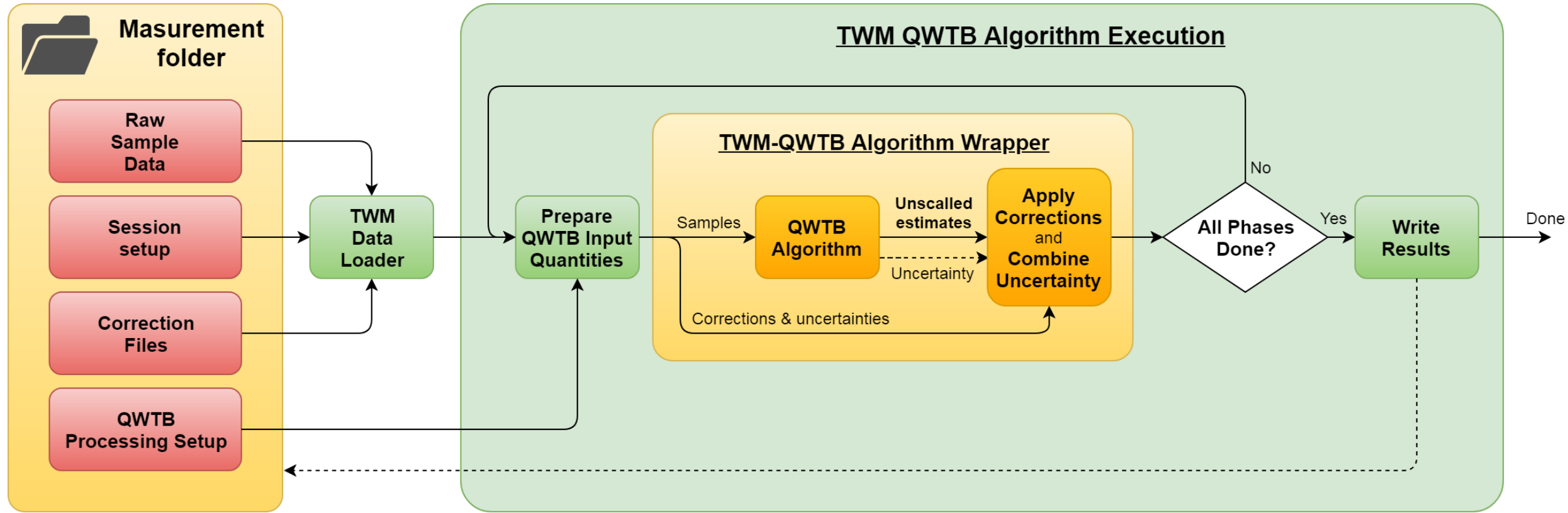
TWM Algorithms

Concept:

- All algorithms are integrated in **QWTB toolbox** developed in Q-WAVE
<https://qwtb.github.io/qwtb/>
 - **Fully transparent** (mostly open source)
 - Compatible with **Matlab** or **GNU Octave** environments
 - Easy addition of new algorithms
- QWTB Toolbox:
 - Collection of **signal processing algorithms**
 - **Unified format** of Input and Output quantities
 - Contains plain algorithms with **NO CORRECTIONS**



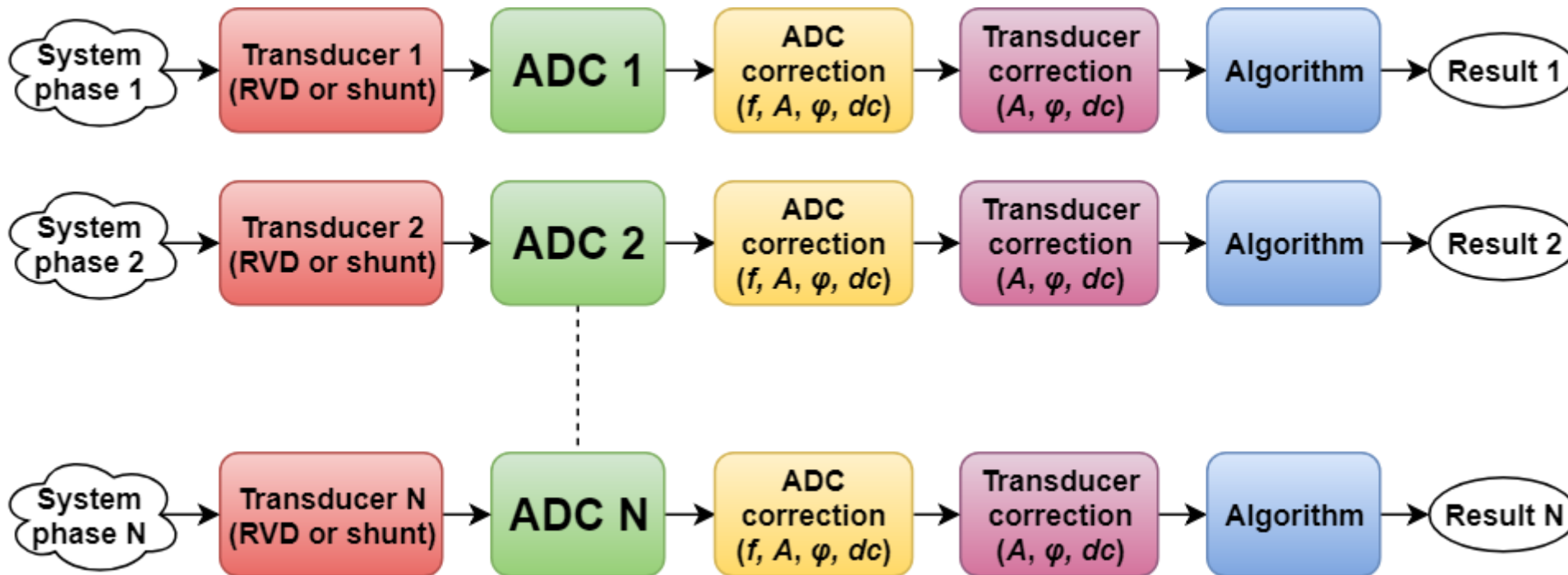
TWM Algorithms



TWM concept – Corrections logic flow

Single channel, multiphase operation:

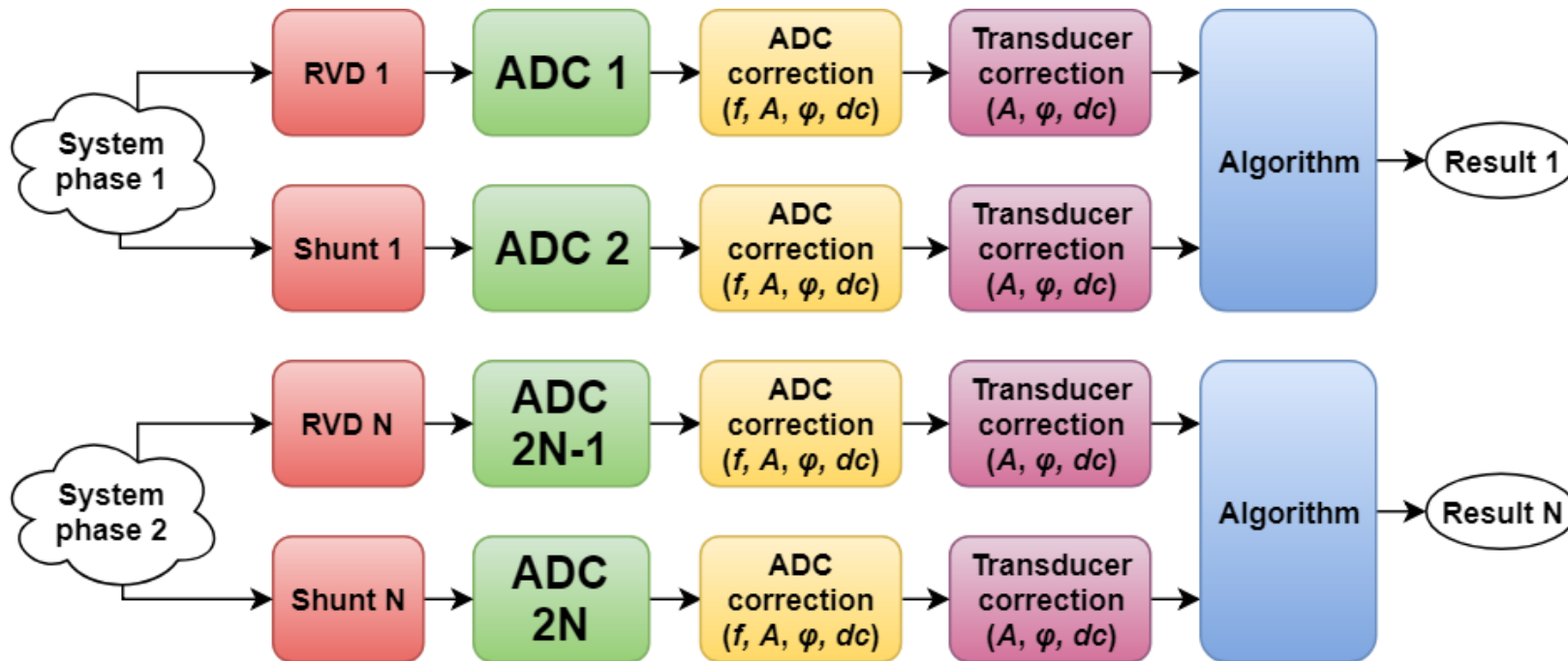
- Each available channel is processed using the same chain of corrections



TWM concept – Corrections logic flow

Multiphase, two input algorithms (power):

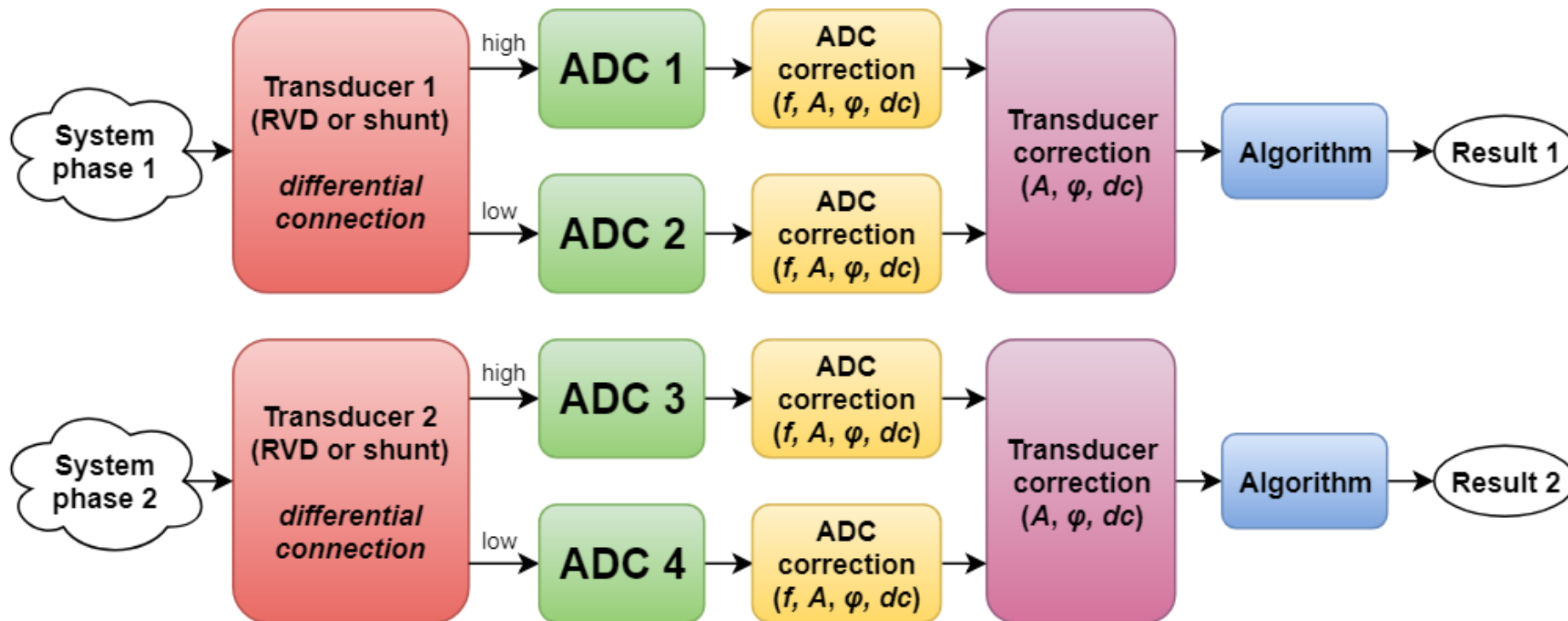
- TWM identifies matching pairs of voltage and current channels for particular phase and applies the same algorithm to all phases



TWM concept – Corrections logic flow

Single channel, multiphase, differential connections:

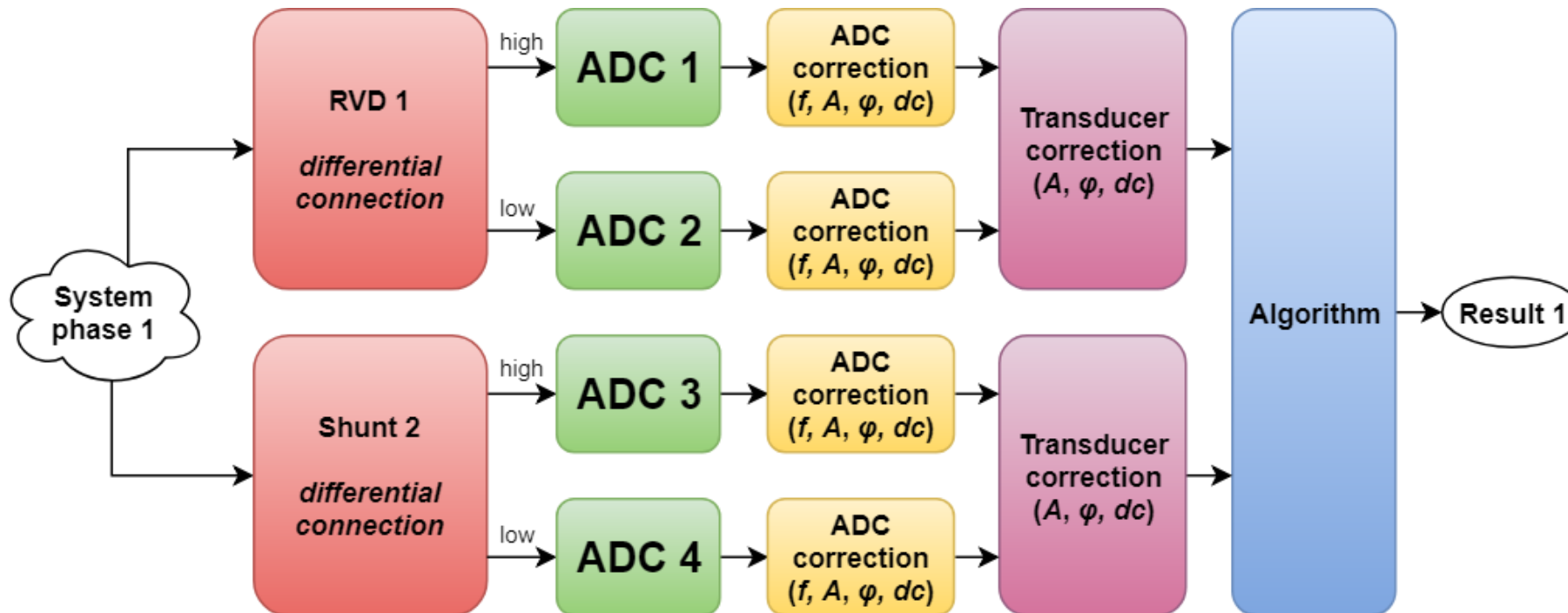
- TWM assigns two channels of digitizer per transducer
- TWM Repeats calculation of selected algorithm for each phase



TWM concept – Corrections logic flow

Multiphase, two input algorithm (power), differential connections:

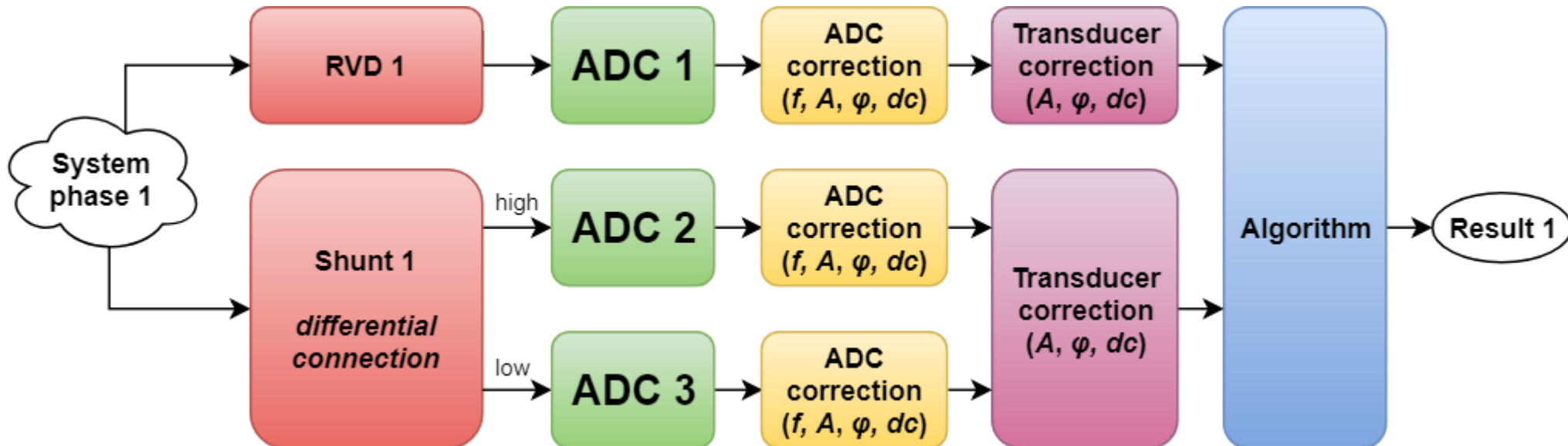
- TWM assigns two channels of digitizer per transducer
- TWM identifies pair of transducers (voltage, current) matching to each phase
- TWM Repeats calculation of selected algorithm for each phase



TWM concept – Corrections logic flow

Multiphase, two input algorithm (power), mixed connections:

- TWM assigns one or two channels of digitizer per transducer (single or differential connection)
- TWM identifies pair of transducers (voltage, current) matching to each phase
- TWM Repeats calculation of selected algorithm for each phase



TWM Algorithms

Concept:

- All algorithms are integrated in QWTB toolbox developed in Q-WAVE
<https://qwtb.github.io/qwtb/>
 - Fully transparent (mostly open source)
 - Compatible with Matlab or GNU Octave environments
 - Easy addition of new algorithms
- TWM contains QWTB with original PLAIN algorithms (*PSFE*, *WFFT*, ...):
 - No corrections and mostly no uncertainty evaluation!
- TWM adds **TWM wrappers of the algorithm** (*TWM-PSFE*, *TWM-WFFT*, *TWM-WRMS*, ...):
 - Implementation of **digitizer** and **transducer corrections** and **uncertainty evaluation**
 - Modifications to cooperate with TWM



TWM Algorithms – General concept

Features:

- Most algorithms for **non-coherent sampling** (no need for synchronization)
- Robust
- Equipped by **uncertainty evaluator**:
 - Monte-Carlo (slow, but accurate)
 - Fast estimator based on previous uncertainty analysis (fast, but lower accuracy)

Groups of algorithms:

- Single-harmonic analysis (Sine fitting, PSFE)
- Multi-harmonic analysis (FFT, Multi-harmonic fitting)
- RMS and Power
- Special PQ (Flicker, AM modulation, Half-cycle RMS, Dip/Swell/Interruption)
- Other (“helper” algorithms)

TWM Algorithms – PSFE

PSFE - **P**hase **S**ensitive **F**requency **E**stimator:

- Estimation of fundamental **F**requency, **A**mplitude and **P**hase angle
- For **non-coherent sampling**
- Immune to harmonic distortion (up to some level)
- TWM wrapper for **single-ended transducers** only
 - Differential mode usable for frequency only
- TWM equipped by fast **uncertainty estimator** for **F**requency only

f, A, φ

Citation:

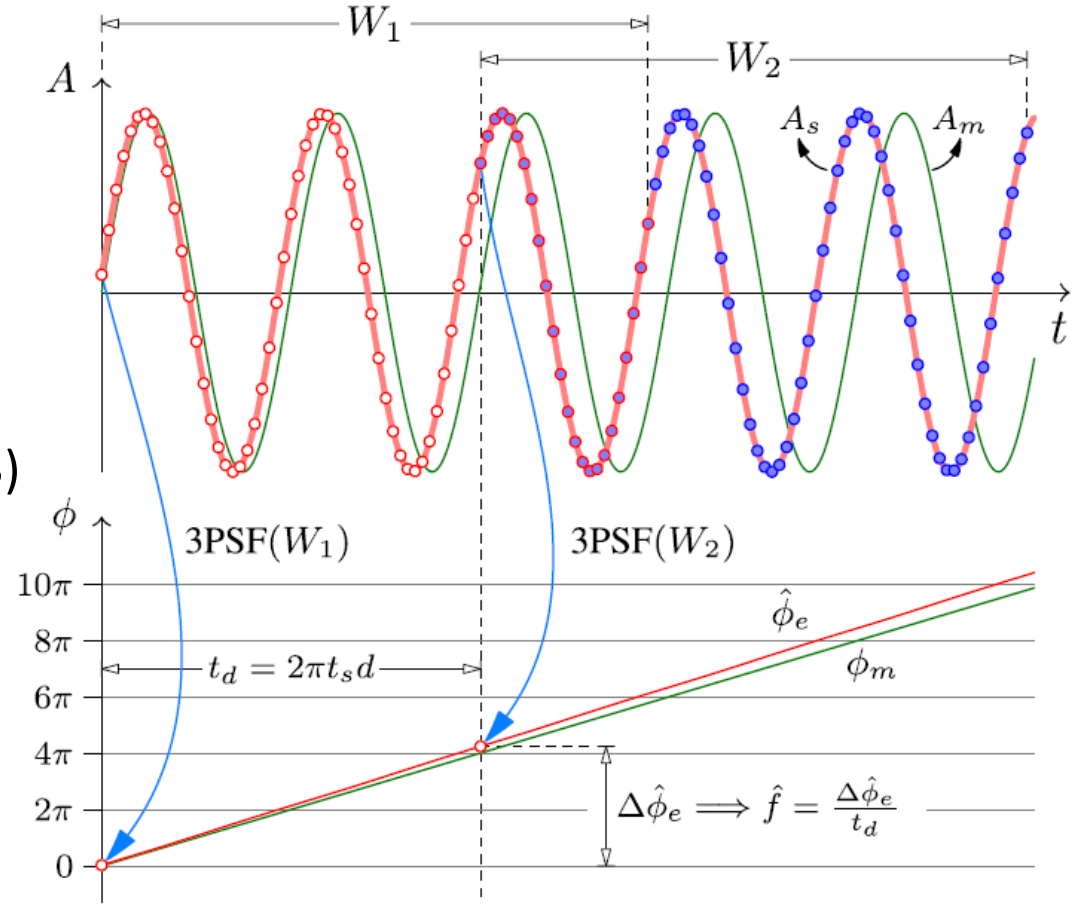
Lapuh, R., "*Estimating the Fundamental Component of Harmonically Distorted Signals From Noncoherently Sampled Data*," Instrumentation and Measurement, IEEE Transactions on , vol.64, no.6, pp.1419,1424, June 2015, doi: 10.1109/TIM.2015.2401211

1. Initial estimate of fundamental frequency

3. Estimation of **phase** of both parts W_1, W_2

using 3-parameter fitting (3PSF)

4. Changing estimate of frequency and repeat from 3) until phase difference W_1 and W_2 is minimized



TWM Algorithms – FPNLSF (4-Param. Non Linear Sine Fit)

FPNLSF – Four Parameter Non Linear Sine Fit:

- Fitting of waveform by 4-parameter model: $y_m = o + A \cdot \sin(2\pi \cdot f \cdot t + \varphi)$
- Returns Frequency, Amplitude, Phase angle and DC offset
- For non-coherent sampling
- Highly sensitive to harmonic distortion – only for pure signals
- Needs initial estimate of frequency accurate to 500 ppm
- Needs as much samples per period as possible (ideally > 100)
- TWM wrapper for single-ended and differential transducers
- TWM equipped by fast uncertainty estimator

f, A, φ, DC

Citation:

M. Šíra and S. Mašláň, "Uncertainty analysis of non-coherent sampling phase meter with four parameter sine wave fitting by means of Monte Carlo," 29th Conference on Precision Electromagnetic Measurements (CPEM 2014), Rio de Janeiro, 2014, pp. 334-335. doi: 10.1109/CPEM.2014.6898395

TWM Algorithms – WFFT (Windowed FFT)

WFFT – **W**indowed **F**ast **F**ourier **T**ransform:

- FFT with optional windowing function
- Returns harmonic **Amplitude(s)**, **Phase(s)**, **DC offset** and RMS estimate
- Designed for **coherent sampling**:
 - Good choice for **inter-channel phase error** measurement and **amplitude ratio** measurement
- Usable for **non-coherent in special case**:
 - **Window functions** other than rectangular (e.g. Blackman)
 - Needs sufficient **harmonic spacing** in the spectrum, e.g. 5 DFT bins for Blackman! (to avoid spectral leakage problems)
 - Measurement **amplitude ratio**, **phase difference**
 - **Not usable for absolute amplitude measurement!**
- TWM wrapper for **single-ended** and **differential transducers**
- TWM equipped by fast **uncertainty estimator**

A, φ , DC



CLASSIC

TWM Algorithms – THDWFFT (THD by Windowed FFT)

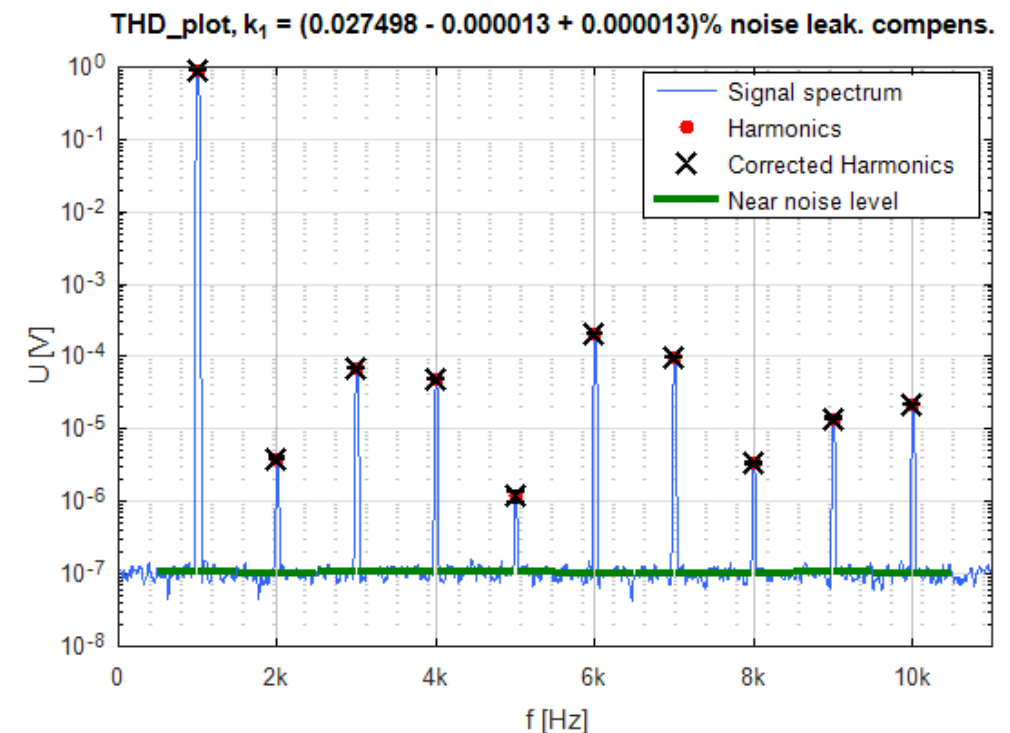
THDWFFT – **T**otal **H**armonic **D**istortion by **W**indowed **F**ast **F**ourier **T**ransform:

A, THD

- Returns **Fundamental frequency**, **Amplitude(s)** and **THD estimates**
- Additionally returns **RMS noise** estimate and **THD+Noise** estimates (no uncertainty)
- Designed for **non-coherent sampling**
- Uses wide window Flattop 248D:
 - **Errors** of amplitude estimates **down to 50 ppm**
 - Requires harmonic spacing **at least 30 DFT bins**
- Initially designed for low-THD measurement
- TWM wrapper for **single-ended** transducers only
- TWM equipped by fast **Monte Carlo calculator**

Citation:

S. Mašláň and M. Šíra, "Automated non-coherent sampling THD meter with spectrum analyser," 29th Conference on Precision Electromagnetic Measurements (CPEM 2014), Rio de Janeiro, 2014, pp. 488-489.
doi: 10.1109/CPEM.2014.6898472



TWM Algorithms – MFSF (Multi Frequency Sine Fitting)

MFSF – Multi Frequency Sine wave Fitting:

f, A, φ, DC
 THD

- Fitting of signal by multi-harmonic model, non-coherent sampling
- Should produce no bias when all harmonics in the signal fitted
- Can fit inter-harmonics if frequency ratios are known in advance
- Returns Fundamental frequency, Amplitude(s), Phase(s), DC offset and THD estimate
- TWM wrapper for single-ended transducers only
- TWM equipped by fast uncertainty estimator and Monte Carlo calculator

Citation:

J. Schoukens, R. Pintelon, and G. Vandersteen, "A sinewave fitting procedure for characterizing data acquisition channels in the presence of time base distortion and time jitter," IEEE Trans. Instrum. Meas., Vol. 46, No. 4, Aug. 1997, 1005-1010

R. Lapuh, "Sampling with 3458A", Left Right d.o.o., September 2018, ISBN 978-961-94476-0-4

TWM Algorithms – WRMS (Windowed RMS)

WRMS – **W**indowed **R**oot **M**ean **S**quare:

- Traditional RMS algorithm **improved** for better immunity to **non-coherent sampling**:

RMS, DC

$$rms = \sqrt{\frac{\sum_{k=1}^N y(k)^2 w(k)^2}{\sum_{k=1}^N w(k)^2}}, \text{ where } w() \text{ is window function and } N \text{ number of samples}$$

- Returns **RMS** and **DC offset**
- Uses Blackman-Harris window:
 - **Error** of the algorithm drops below **ppm level** when **at least 8 periods** of signals recorded
- Operates in time-domain
- TWM wrapper for **single-ended** and **differential transducers**
- TWM equipped by fast **uncertainty estimator** and **Monte Carlo**

Citation:

R. Lapuh, B. Voljc, M. Kokalj, B. Pinter, Z. Svetik and M. Lindic, "*Measurement of repetitive arbitrary waveform RMS value*," 29th Conference on Precision Electromagnetic Measurements (CPEM 2014), Rio de Janeiro, 2014, pp. 420-421. doi: 10.1109/CPEM.2014.6898438

TWM Algorithms – PWRTDI (Windowed RMS Power)

PWRTDI – **Power** using **Time Domain Integration**:

- Extension of traditional power in time-domain for **non-coherent sampling**:

$$P = \sqrt{\frac{\sum_{k=1}^N u(k)i(k)w(k)^2}{\sum_{k=1}^N w(k)^2}}, \text{ where } w() \text{ is window function and } N \text{ number of samples}$$

- Returns **P, S, Q, PF, U_{RMS}, I_{RMS}, DC power, DC U, DC I**
- Works in all **4 quadrants** – detects quadrant using auxiliary FFT analysis:
 - Quadrant detection may fail for low power factors
- Uses Blackman-Harris window:
 - **Error** of the algorithm drops to **ppm level** when **at least 30 periods** of signals recorded
- TWM wrapper for **single-ended** and **differential transducers**
- TWM equipped by fast **uncertainty estimator** and **Monte Carlo**

RMS, DC
P, Q, S, PF

Citation:

K. B. Ellingsberg, “Predictable maximum RMS-error for windowed RMS (RMWS),” 2012 Conference on Precision electromagnetic Measurements, Washington, DC, 2012, pp. 308-309. doi: 10.1109/CPEM.2012.625092

TWM Algorithms – PWRTDI (Windowed RMS Power)

PWRTDI – Definitions of quantities:

Table 33: Definitions of the TWM-PWRTDI output quantities based on the basic quantities U , I , P , U_{dc} and I_{dc} . Note the input quantities marked * are obtained from the other output quantity.

Returned quantity	AC coupled definition	DC coupled definition
DC voltage U_{dc}	U_{dc}	U_{dc}
DC current I_{dc}	I_{dc}	I_{dc}
DC power component P_{dc}	$U_{dc} \cdot I_{dc}$	$U_{dc} \cdot I_{dc}$
RMS voltage U	U	$\sqrt{U^2 + U_{dc}^2}$
RMS current I	I	$\sqrt{I^2 + I_{dc}^2}$
Active power P	P	$P + U_{dc} \cdot I_{dc}$
Reactive power Q	$\sqrt{(UI)^2 - P^2} \cdot \text{sign}(Q_{\text{bud}})$	$\sqrt{(UI)^2 - P^2} \cdot \text{sign}(Q_{\text{bud}})$
Apparent power S	$(UI)^2$	$\sqrt{U^2 + U_{dc}^2} \cdot \sqrt{I^2 + I_{dc}^2}$
Power factor PF	P/S^*	P/S^*
Effective phase angle phi_{ef}	$\text{atan2}(Q^*, P)$	$\text{atan2}(Q^*, P)$

Budenau's definition of reactive power:

$$Q_{\text{bud}} = \sum_{h=1}^H (0.5 \cdot U(h) \cdot I(h) \cdot \sin \phi(h)),$$

TWM Algorithms – PWRFFT (FFT-based Power)

PWRFFT – Power using FFT:

- Power algorithm for coherent sampling
- Returns P , S , Q , PF , U_{RMS} , I_{RMS} , DC power, DC U, DC I
- Works in all 4 quadrants
- AC/DC coupling

- TWM wrapper for single-ended and differential transducers
- TWM equipped by fast uncertainty estimator

RMS, DC
P, Q, S, PF



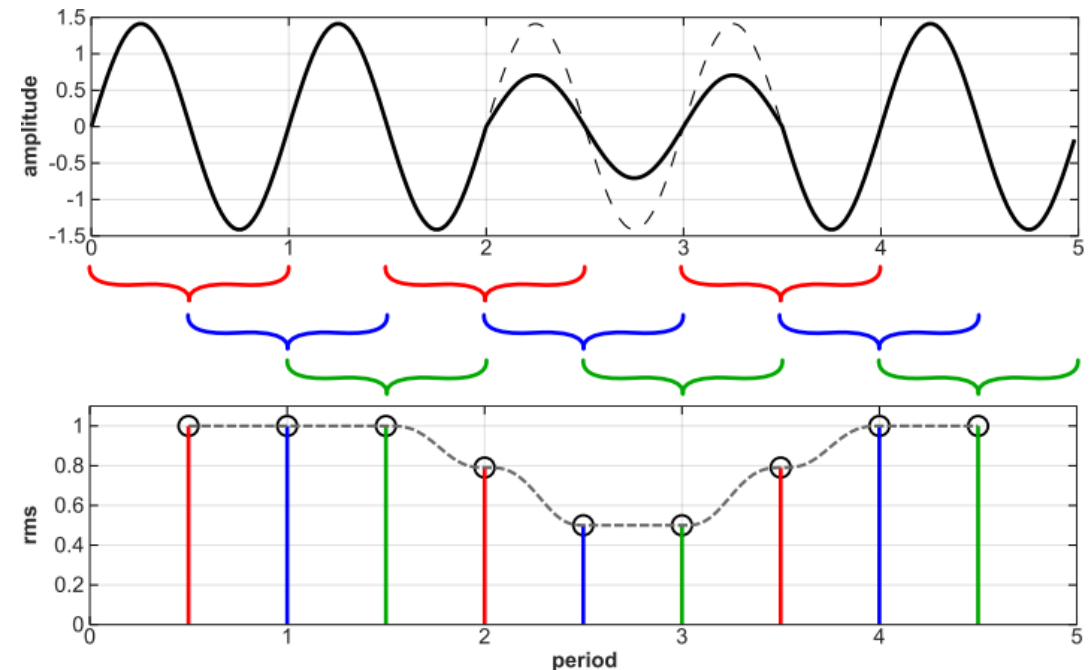
CLASSIC

TWM Algorithms – HCRMS (Half-Cycle RMS)

HCRMS – Half-Cycle RMS:

HC-RMS

- Calculation of so called “Half Cycle” RMS value according IEC 61000-4-30
 - A class (half-cycle step) or S class (sliding window)
- Designed for non-coherent sampling (resamples the signal to coherent internally)
- Returns RMS level envelope as a function of time
- Usable to several PQ tests:
 - Mainly Dip, Swell, Interruption
- Requires as many as possible samples per period
- TWM wrapper for single-ended transducers only
- TWM equipped by fast uncertainty estimator

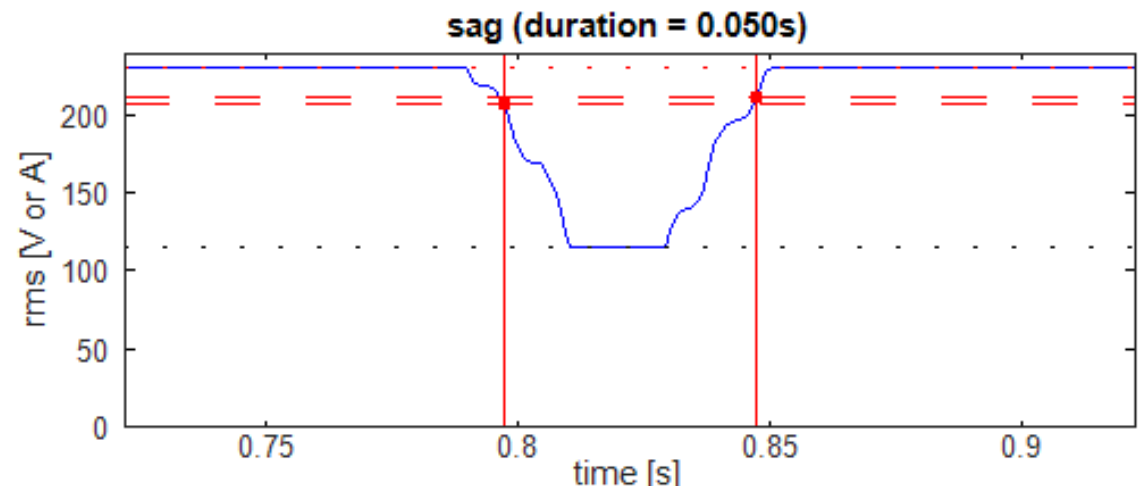


TWM Algorithms – InDiSwell (Interrupt, Dip, Swell)

InDiSwell – **Interrupt Dip Swell** detector:

- Detection of event according **IEC 61000-4-30** in **A class** (half-cycle step) or **S class** (sliding window):
 - Dip (Sag) – decrease of RMS
 - Swell – increase of RMS
 - Interruption – severe decrease of RMS (below 10% of nominal)
- Designed for **non-coherent sampling** (resamples the signal to coherent internally)
- Returns **Duration of event**, **Relative event start time**, **Residual level**
- Configurable **Thresholds**, **Hysteresis**
- TWM wrapper for **single-ended** transducers only
- TWM equipped by fast **uncertainty estimator**

HC-RMS
Dip, Swell
Interruption



TWM Algorithms – Flicker

Flicker – **Flicker** detector:

- Detection of P_{ST} according IEC 61000-4-15
- Designed for non-coherent sampling
- Returns P_{INST} , P_{ST}
- Automatic resampling to optimal sampling rate
- **Warning - Slightly different results in Matlab and GNU Octave**
- Requires long record – at least 11 minutes of continuous record with sampling rate higher than 7 kSa/s
- TWM wrapper for single-ended transducers only
- TWM equipped by fast uncertainty estimator

P_{st} , P_{inst}

Citations:

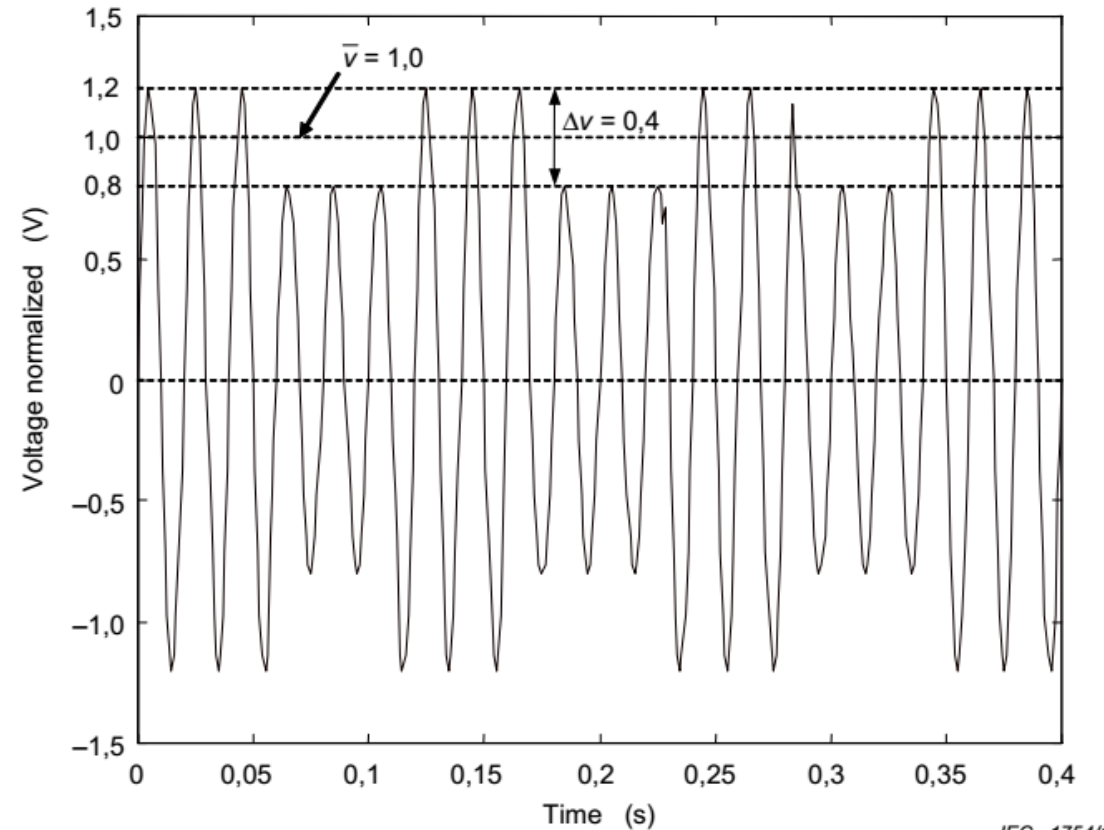
- IEC 61000-4-15, Electromagnetic compatibility (EMC), Testing and measurement techniques, Flickermeter, Edition 2.0, 2010-08
- Wilhelm Mombauer: "Messung von Spannungsschwankungen und Flickern mit dem IEC-Flickermeter", ISBN 3-8007-2525-8, VDE-Verlag
- Solcept Open Source Flicker Measurement-Simulator <https://www.solcept.ch/en/tools/flickersim/>
- NPL Reference Flickermeter Design <http://www.npl.co.uk/electromagnetics/electrical-measurement/products-and-services/npl-reference-flickermeter-design>

TWM Algorithms – Flicker

Definitions:

- P_{ST} – Short Term Flicker Severity:
 - Expresses human perceptibility to flickering light source
- $\Delta V/V$ (or $\Delta U/U$) – modulation depth
- CPM – Changes Per Minute:
 - Different expression of modulating frequency f_M :
 $CPM = 120 \cdot f_M$

Definition of $\Delta V/V$:



IEC 1754/10

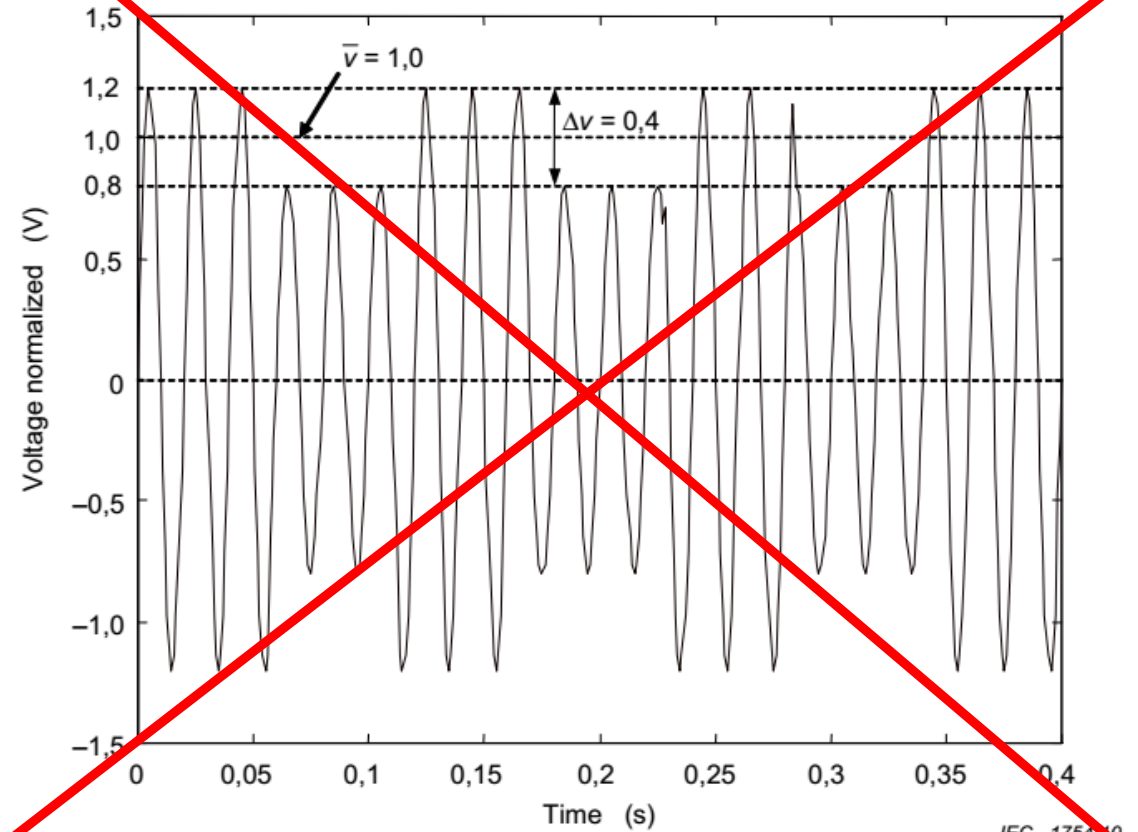
Source: IEC 61000-4-15

TWM Algorithms – Flicker

Definitions:

- P_{ST} – Short Term Flicker Severity:
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- CPM – Changes Per Minute:
 - Different expression of modulating frequency f_M :
 $CPM = 120 \cdot f_M$

Definition of $\Delta V/V$:

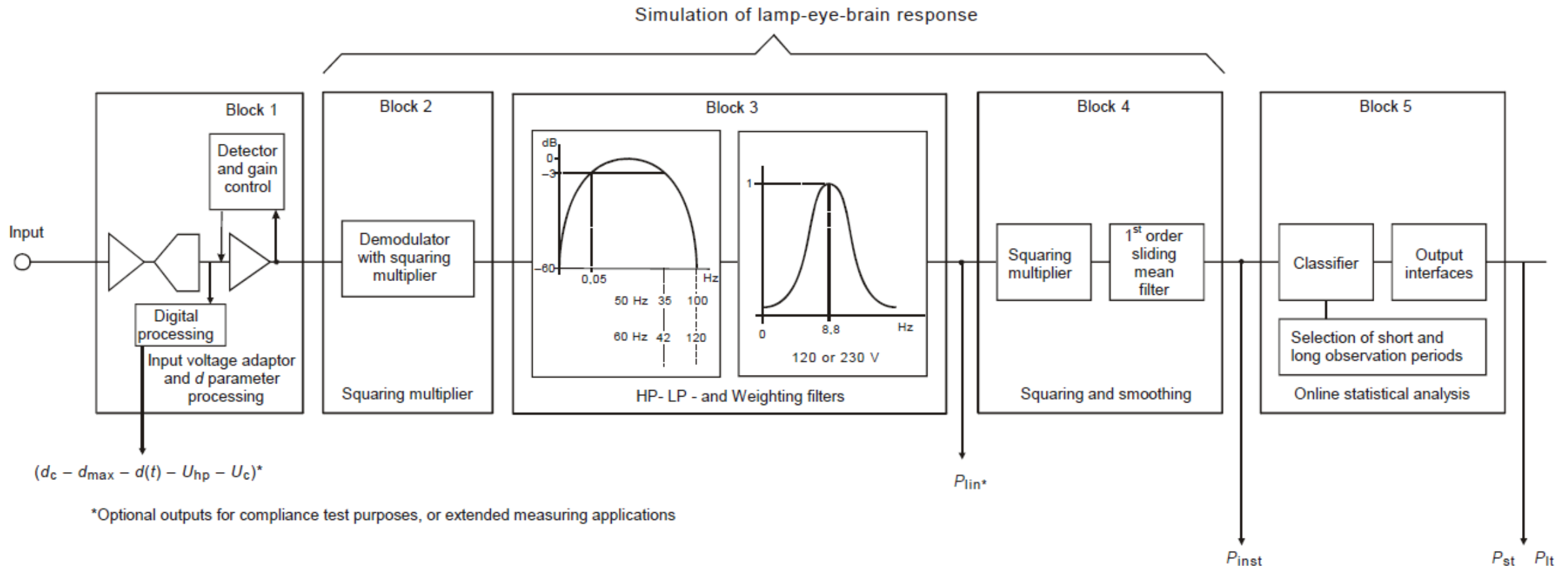


IEC 1754:0

Source: IEC 61000-4-15

TWM Algorithms – Flicker

Flicker calculation according **IEC 61000-4-15**:



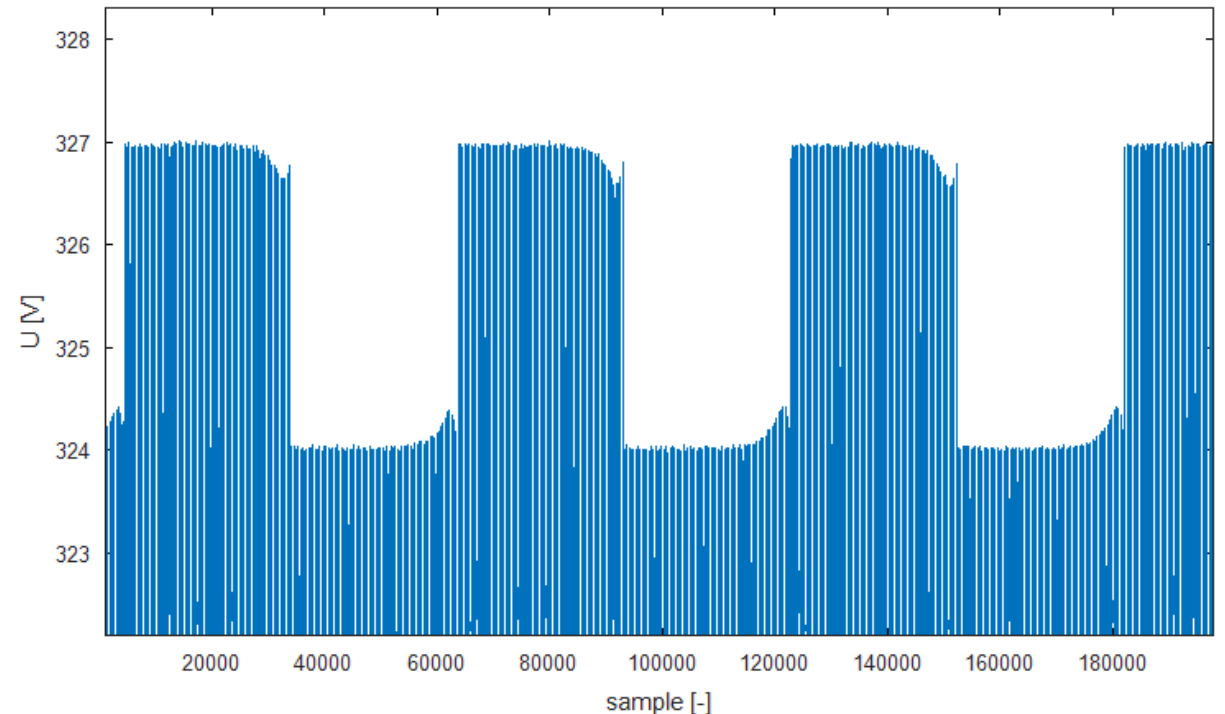
Source: IEC 61000-4-15

TWM Algorithms – Flicker

Hint for calibration:

- Manufacturers of calibrators uses two approaches:
 - Define CPM, dVV and shape of modulation, optionally shows user corresponding P_{ST} value
 - Select preset value of P_{ST} from a list
- Manufacturer may expect ideal shape:
 - Not always the case (see example)
 - The displayed value of P_{ST} may be incorrect because it is calculated from expected shape
 - Rather evaluate CPM and dVV instead
- See waveform whenever the measured P_{ST} does not match the set value

Actual measurement of Flicker modulation on calibrator:



TWM Algorithms – MODTDPS (AM modulation estimator)

MODTDPS – AM **Mod**ulation from quadrature **Time Domain Phase Shift**:

- Measurement of amplitude modulation parameters for **sine** or **rectangular modulation** of **sine carrier**
- Alternative to Flicker algorithm when evaluating just $\Delta V/V$ and **CPM**
- Designed for **non-coherent sampling**
- Based on complex **analytical signal**:

$$\hat{y}_m(t) = y(t) + j \cdot y\left(t + \frac{\pi}{2 \cdot f_0}\right), \text{ where } f_0 \text{ is carrier frequency}$$

- Returns: **Carrier frequency, Carrier amplitude, Modulating frequency, Modulating amplitude, Modulation envelope, dVV and CPM**

- TWM wrapper for **single-ended** and **differential** transducers
- TWM equipped by fast **uncertainty estimator** only for single-ended case

$\Delta V/V$, CPM
 f_0, A_0, f_m, A_m